

bq25601 I²C Controlled 3-A, Single-Cell Battery Charger for High Input Voltage and Narrow Voltage DC (NVDC) Power Path Management

1 Features

- High-Efficiency, 1.5-MHz, Synchronous Switch-Mode Buck Charger
 - 92% Charge Efficiency at 2 A from 5-V Input
 - Optimized for USB Voltage Input (5 V)
 - Selectable Low Power Pulse Frequency Modulation (PFM) Mode for Light Load Operations
- Supports USB On-The-Go (OTG)
 - Boost Converter With Up to 1.2-A Output
 - 92% Boost Efficiency at 1-A Output
 - Accurate Constant Current (CC) Limit
 - Soft-Start Up To 500- μ F Capacitive Load
 - Output Short Circuit Protection
 - Selectable Low Power PFM Mode for Light Load Operations
- Single Input to Support USB Input and High Voltage Adaptors
 - Support 3.9-V to 13.5-V Input Voltage Range With 20-V Absolute Maximum Input Voltage Rating
 - Programmable Input Current Limit (100 mA to 3.2 A With 100-mA Resolution) to Support USB 2.0, USB 3.0 Standards and High Voltage Adaptors (IINDPM)
 - Maximum Power Tracking by Input Voltage Limit Up to 5.4 V (VINDPM)
 - VINDPM Threshold Automatically Tracks Battery Voltage
- High Battery Discharge Efficiency With 18-m Ω Battery Discharge MOSFET
- Narrow VDC (NVDC) Power Path Management
 - Instant-On Works with No Battery or Deeply Discharged Battery
 - Ideal Diode Operation in Battery Supplement Mode
- BATFET Control to Support Ship Mode, Wake Up and Full System Reset
- Flexible Autonomous and I²C Mode for Optimal System Performance
- High Integration Includes All MOSFETs, Current Sensing and Loop Compensation
- 17- μ A Low Battery Leakage Current to Support Ship Mode
- High Accuracy

- \pm 0.5% Charge Voltage Regulation
- \pm 5% at 1.5-A Charge Current Regulation
- \pm 10% at 0.9-A Input Current Regulation
- Safety
 - Battery Temperature Sensing for Charge and Boost Mode
 - Thermal Regulation and Thermal Shutdown
 - Input UVLO and Overvoltage Protection

2 Applications

- Smart Phone, ePOS
- Portable Internet Devices and Accessory

3 Description

The bq25601 device is a highly-integrated 3-A switch-mode battery charge management and system power path management device for single cell Li-Ion and Li-polymer battery. The low impedance power path optimizes switch-mode operation efficiency, reduces battery charging time and extends battery life during discharging phase. The I²C serial interface with charging and system settings makes the device a truly flexible solution.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
bq25601	WQFN (24)	4.00 mm x 4.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Simplified Schematic

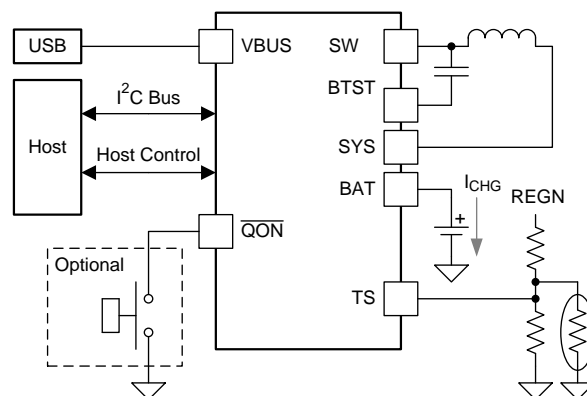


Table of Contents

1 Features	1	8.4 Device Functional Modes	17
2 Applications	1	8.5 Register Maps	33
3 Description	1	9 Application and Implementation	44
4 Revision History	2	9.1 Application information	44
5 Description (continued)	2	9.2 Typical Application Diagram	45
6 Pin Configuration and Functions	4	9.3 Application Curves	46
7 Specifications	6	10 Power Supply Recommendations	50
7.1 Absolute Maximum Ratings	6	11 Layout	51
7.2 Recommended Operating Conditions	6	11.1 Layout Guidelines	51
7.3 Thermal information	7	11.2 Layout Example	51
7.4 Electrical Characteristics	7	12 Device and Documentation Support	53
7.5 Timing Requirements	12	12.1 Community Resources	53
7.6 Typical Characteristics	13	12.2 Trademarks	53
8 Detailed Description	15	12.3 Electrostatic Discharge Caution	53
8.1 Overview	15	12.4 Glossary	53
8.2 Functional Block Diagram	16	13 Mechanical, Packaging, and Orderable Information	54
8.3 Feature Description	16		

4 Revision History

DATE	REVISION	NOTES
March 2017	*	Initial release.

5 Description (continued)

The bq25601 is a highly-integrated 3.0-A switch-mode battery charge management and system power path management device for single cell Li-Ion and Li-polymer battery. It features fast charging with high input voltage support for a wide range of smart phones, tablets and portable devices. Its low impedance power path optimizes switch-mode operation efficiency, reduces battery charging time and extends battery life during discharging phase. Its input voltage and current regulation deliver maximum charging power to battery. The solution is highly integrated with input reverse-blocking FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET, Q4) between system and battery. It also integrates the bootstrap diode for the high-side gate drive for simplified system design. The I²C serial interface with charging and system settings makes the device a truly flexible solution.

The device supports a wide range of input sources, including standard USB host port, USB charging port, and USB compliant high voltage adapter. To support fast charging using high voltage adapter, the bq25601 provides PSEL pin for USB switch control. To set the default input current limit, device takes the result from detection circuit in the system, such as USB PHY device. The device is compliant with USB 2.0 and USB 3.0 power spec with input current and voltage regulation. The device also meets USB On-the-Go (OTG) operation power rating specification by supplying 5.15 V on VBUS with constant current limit up to 1.2-A.

The power path management regulates the system slightly above battery voltage but does not drop below 3.5 V minimum system voltage (programmable). With this feature, the system maintains operation even when the battery is completely depleted or removed. When the input current limit or voltage limit is reached, the power path management automatically reduces the charge current to zero. As the system load continues to increase, the power path discharges the battery until the system power requirement is met. This Supplement Mode prevents overloading the input source.

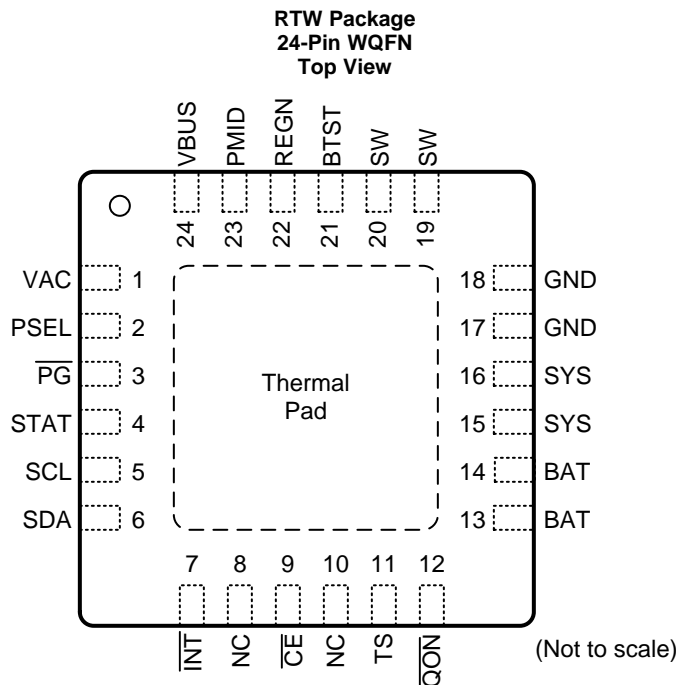
The device initiates and completes a charging cycle without software control. It automatically detects the battery voltage and charges the battery in three phases: pre-conditioning, constant current and constant voltage. At the end of the charging cycle, the charger automatically terminates when the charge current is below a preset limit in the constant voltage phase. When the full battery falls below the recharge threshold, the charger automatically starts another charging cycle.

The charger provides various safety features for battery charging and system operations, including battery temperature negative thermistor monitoring, charging safety timer and overvoltage and overcurrent protections. The thermal regulation reduces charge current when the junction temperature exceeds 110°C (programmable). The STAT output reports the charging status and any fault conditions. The VBUS_GD bit indicates if a good power source is present. The INT output immediately notifies host when fault occurs.

The device also provides $\overline{\text{QON}}$ pin for BATFET enable and reset control to exit low power ship mode or full system reset function.

The device is available in 24-pin, 4 mm x 4 mm x 0.75 mm thin WQFN package.

6 Pin Configuration and Functions



Pin Functions

Pin		TYPE ⁽¹⁾	DESCRIPTION
NAME	NO.		
BAT	13	P	Battery connection point to the positive terminal of the battery pack. The internal BATFET and current sensing is connected between SYS and BAT. Connect a 10 μ F close to the BAT pin.
	14		
BTST	21	P	PWM high side driver positive supply. Internally, the BTST pin is connected to the cathode of the boost-strap diode. Connect the 0.047- μ F bootstrap capacitor from SW to BTST.
\overline{CE}	9	DI	Charge enable pin. When this pin is driven low, battery charging is enabled.
GND	17	—	Ground pins.
	18		
\overline{INT}	7	DO	Open-drain interrupt Output. Connect the INT to a logic rail through 10-k Ω resistor. The \overline{INT} pin sends an active low, 256- μ s pulse to host to report charger device status and fault.
NC	8	—	No Connect. Keep the pins float.
	10		
\overline{PG}	3	DO	Open drain active low power good indicator. Connect to the pull up rail through 10-k Ω resistor. LOW indicates a good input source if the input voltage is between UVLO and ACOV, above SLEEP mode threshold, and current limit is above 30 mA.
PMID	23	DO	Connected to the drain of the reverse blocking MOSFET (RBFET) and the drain of HSFET. Put 10 μ F ceramic capacitor on PMID to GND.
PSEL	2	DI	Power source selection input. Set 500 mA input current limit by pulling this pin high and set 2.4A input current limit by pulling this pin low. Once the device gets into host mode, the host can program different input current limits to IINDPM register.
\overline{QON}	12	DI	BATFET enable/reset control input. When BATFET is in ship mode, a logic low of $t_{SHIPMODE}$ duration turns on BATFET to exit shipping mode. When VBUS is not plugged-in, a logic low of t_{QON_RST} (minimum 8 s) duration resets SYS (system power) by turning BATFET off for t_{BATFET_RST} (minimum 250 ms) and then re-enable BATFET to provide full system power reset. The pin contains an internal pull-up to maintain default high logic.
REGN	22	P	LSFET driver and internal supply output. Internally, REGN is connected to the anode of the boost-strap diode. Connect a 4.7- μ F (10-V rating) ceramic capacitor from REGN to GND. The capacitor should be placed close to the IC.
SCL	5	DI	I ² C interface clock. Connect SCL to the logic rail through a 10-k Ω resistor.
SDA	6	DIO	I ² C interface data. Connect SDA to the logic rail through a 10-k Ω resistor.

(1) AI = Analog input, AO = Analog Output, AIO = Analog input Output, DI = Digital input, DO = Digital Output, DIO = Digital input Output, P = Power

Pin Functions (continued)

Pin		TYPE ⁽¹⁾	DESCRIPTION
NAME	NO.		
STAT	4	DO	Open-drain charge status output. Connect the STAT pin to a logic rail via 10-k Ω resistor. The STAT pin indicates charger status. Collect a current limit resistor and a LED from a rail to this pin. Charge in progress: LOW Charge complete or charger in SLEEP mode: HIGH Charge suspend (fault response): 1-Hz, 50% duty cycle Pulses This pin can be disabled via EN_ICHG_MON[1:0] register bits.
SW	19	P	Switching node output. Connected to output inductor. Connect the 0.047- μ F bootstrap capacitor from SW to BTST.
	20		
SYS	15	P	Converter output connection point. The internal current sensing network is connected between SYS and BAT. Connect a 20 μ F capacitor close to the SYS pin.
	16		
TS	11	AI	Temperature qualification voltage input to support JEITA profile. Connect a negative temperature coefficient thermistor. Program temperature window with a resistor divider from REGN to TS to GND. Charge suspends when either TS pin is out of range. When TS pin is not used, connect a 10-k Ω resistor from REGN to TS and connect a 10-k Ω resistor from TS to GND. It is recommended to use a 103AT-2 thermistor.
VAC	1	AI	Charge input voltage sense. This pin must be connected to VBUS pin.
VBUS	24	P	Charger input. The internal n-channel reverse block MOSFET (RBFET) is connected between VBUS and PMID pins. Place a 1- μ F ceramic capacitor from VBUS to GND close to device.
Thermal Pad	—	P	Thermal pad and ground reference. This pad is ground reference for the device and it is also the thermal pad used to conduct heat from the device. This pad should be tied externally to a ground plane through PCB vias under the pad.

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		Min	MAX	UNIT
Voltage Range (with respect to GND)	VAC, VBUS (converter not switching)	-2	20	V
Voltage Range (with respect to GND)	BTST, PMID (converter not switching)	-0.3	20	V
Voltage Range (with respect to GND)	SW	-2	16	V
Voltage Range (with respect to GND)	BTST to SW	-0.3	7	V
Voltage Range (with respect to GND)	PSEL	-0.3	7	V
Voltage Range (with respect to GND)	REGN, TS, \overline{CE} , \overline{PG} , BAT, SYS (converter not switching)	-0.3	7	V
Output Sink current	STAT		6	mA
Voltage Range (with respect to GND)	SDA, SCL, INT, /QON, STAT	-0.3	7	V
Voltage Range (with respect to GND)	PGND to GND (QFN package only)	-0.3	0.3	V
Output Sink current	INT		6	mA
Operating junction temperature, T _J		-40	150	°C
Storage temperature, T _{stg}		-65	150	°C

- (1) Stresses beyond those listed under Absolute maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. All voltage values are with respect to the network ground terminal unless otherwise noted.

7.2 Recommended Operating Conditions

		Min	NOM	MAX	UNIT
V _{BUS}	input voltage	3.9		13.5 ⁽¹⁾	V
I _{in}	Input current (VBUS)			3.25	A
I _{SWOP}	Output current (SW)			3.25	A
V _{BATOP}	Battery voltage			4.624	V
I _{BATOP}	Fast charging current			3.0	A
I _{BATOP}	Discharging current (continuous)			6	A
T _A	Operating ambient temperature	-40		85	°C

- (1) The inherent switching noise voltage spikes should not exceed the absolute maximum voltage rating on either the BTST or SW pins. A tight layout minimizes switching noise.

7.3 Thermal information

THERMAL METRIC ⁽¹⁾		bq25601	
		RTW (WQFN)	UNIT
		24 PinS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	35.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	22.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	11.9	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	0.2	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	12	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	2.6	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

7.4 Electrical Characteristics

$V_{VAC_UVLOZ} < V_{VAC} < V_{VAC_OV}$ and $V_{VAC} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to 125°C and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
QUIESCENT CURRENTS						
I_{BAT}	Battery discharge current (BAT, SW, SYS) in buck mode	$V_{BAT} = 4.5\text{ V}$, $V_{BUS} < V_{VAC_UVLOZ}$, leakage between BAT and VBUS, $T_J < 85^{\circ}\text{C}$		5	μA	
I_{BAT}	Battery discharge current (BAT) in buck mode	$V_{BAT} = 4.5\text{ V}$, HIZ Mode and OVPFET_DIS = 1 or No VBUS, I2C disabled, BATFET Disabled. $T_J < 85^{\circ}\text{C}$	17	33	μA	
I_{BAT}	Battery discharge current (BAT, SW, SYS)	$V_{BAT} = 4.5\text{ V}$, HIZ Mode and OVPFET_DIS = 1 or No VBUS, I2C Disabled, BATFET Enabled. $T_J < 85^{\circ}\text{C}$	58	85	μA	
I_{VBUS_HIZ}	Input supply current (VBUS) in buck mode	$V_{VBUS} = 5\text{ V}$, High-Z Mode and OVPFET_DIS = 1, No battery	37	50	μA	
I_{VBUS_HIZ}	Input supply current (VBUS) in buck mode	$V_{VBUS} = 12\text{ V}$, High-Z Mode and OVPFET_DIS = 1, No battery	68	90	μA	
I_{VBUS}	Input supply current (VBUS) in buck mode	$V_{VBUS} = 12\text{ V}$, $V_{VBUS} > V_{VBAT}$, converter not switching	1.5	3	mA	
I_{VBUS}	Input supply current (VBUS) in buck mode	$V_{VBUS} > V_{UVLO}$, $V_{VBUS} > V_{VBAT}$, converter switching, $V_{BAT} = 3.8\text{ V}$, $I_{SYS} = 0\text{ A}$	3		mA	
I_{BOOST}	Battery Discharge Current in boost mode	$V_{BAT} = 4.2\text{ V}$, boost mode, $I_{VBUS} = 0\text{ A}$, converter switching	3		mA	
VBUS, VAC AND BAT PIN POWER-UP						
V_{BUS_OP}	VBUS operating range	V_{VBUS} rising	3.9	13.5	V	
V_{VAC_UVLOZ}	VBUS for active I ² C, no battery Sense VAC pin voltage	V_{VAC} rising	3.3	3.6	V	
$V_{VAC_UVLOZ_HYS}$	I ² C active hysteresis	V_{AC} falling from above V_{VAC_UVLOZ}	300		mV	
$V_{VAC_PRESENT}$	One of the conditions to turn on REGN	V_{VAC} rising	3.65	3.9	V	
$V_{VAC_PRESENT_HYS}$	One of the conditions to turn on REGN	V_{VAC} falling	500		mV	
V_{SLEEP}	Sleep mode falling threshold	$(V_{VAC} - V_{VBAT})$, $V_{BUSMIN_FALL} \leq V_{BAT} \leq V_{REG}$, VAC falling	15	60	110	mV
V_{SLEEPZ}	Sleep mode rising threshold	$(V_{VAC} - V_{VBAT})$, $V_{BUSMIN_FALL} \leq V_{BAT} \leq V_{REG}$, VAC rising	115	220	340	mV
$V_{VAC_OV_RISE}$	VAC 6.5-V Overvoltage rising threshold	VAC rising; OVP (REG06[7:6]) = '01'	6.1	6.4	6.7	V

Electrical Characteristics (continued)

$V_{VAC_UVLOZ} < V_{VAC} < V_{VAC_OV}$ and $V_{VAC} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to 125°C and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{VAC_OV_RISE}$	VAC 10.5-V Overvoltage rising threshold	VAC rising, OVP (REG06[7:6]) = '10'	10.35	10.9	11.5	V
$V_{VAC_OV_RISE}$	VAC 14-V Overvoltage rising threshold	VAC rising, OVP (REG06[7:6]) = '11'	13.5	14.2	14.85	V
$V_{VAC_OV_HYS}$	VAC 6.5-V Overvoltage hysteresis	VAC falling, OVP (REG06[7:6]) = '01'		320		mV
$V_{VAC_OV_HYS}$	VAC 10.5-V Overvoltage hysteresis	VAC falling, OVP (REG06[7:6]) = '10'		250		mV
$V_{VAC_OV_HYS}$	VAC 14-V Overvoltage hysteresis	VAC falling, OVP (REG06[7:6]) = '11'		300		mV
V_{BAT_UVLOZ}	BAT for active I ² C, no adapter	V_{BAT} rising	2.5			V
$V_{BAT_DPL_FALL}$	Battery Depletion Threshold	V_{BAT} falling	2.2		2.6	V
$V_{BAT_DPL_RISE}$	Battery Depletion Threshold	V_{BAT} rising	2.35		2.8	V
$V_{BAT_DPL_HYST}$	Battery Depletion rising hysteresis	V_{BAT} rising		180		mV
V_{BUSMIN_FALL}	Bad adapter detection falling threshold	V_{BUS} falling	3.75	3.9	4.0	V
V_{BUSMIN_HYST}	Bad adapter detection hysteresis			80		mV
I_{BADSRC}	Bad adapter detection current source	Sink current from VBUS to GND		30		mA
POWER-PATH						
V_{SYS_MIN}	System regulation voltage	$V_{VBAT} < SYS_MIN[2:0] = 101$, BATFET Disabled (REG07[5] = 1)	3.5	3.68		V
V_{SYS}	System Regulation Voltage	$I_{SYS} = 0\text{ A}$, $V_{VBAT} > V_{SYSMIN}$, $V_{VBAT} = 4.400\text{ V}$, BATFET disabled (REG07[5] = 1)		$V_{BAT} + 50\text{ mV}$		V
V_{SYS_MAX}	Maximum DC system voltage output	$I_{SYS} = 0\text{ A}$, Q4 off, $V_{VBAT} \leq 4.400\text{ V}$, $V_{VBAT} > V_{SYSMIN} = 3.5\text{ V}$	4.4	4.45	4.48	V
$R_{ON(RBFET)}$	Top reverse blocking MOSFET on-resistance between VBUS and PMID - Q1	$-40^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$		45		m Ω
$R_{ON(HSFET)}$	Top switching MOSFET on-resistance between PMID and SW - Q2	$V_{REGN} = 5\text{ V}$, $-40^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$		62		m Ω
$R_{ON(LSFET)}$	Bottom switching MOSFET on-resistance between SW and GND - Q3	$V_{REGN} = 5\text{ V}$, $-40^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$		71		m Ω
V_{FWD}	BATFET forward voltage in supplement mode			30		mV
$R_{ON(BAT-SYS)}$	SYS-BAT MOSFET on-resistance	QFN package, Measured from BAT to SYS, $V_{BAT} = 4.2\text{ V}$, $T_J = 25^{\circ}\text{C}$		19.5	24	m Ω
$R_{ON(BAT-SYS)}$	SYS-BAT MOSFET on-resistance	QFN package, Measured from BAT to SYS, $V_{BAT} = 4.2\text{ V}$, $T_J = -40 - 125^{\circ}\text{C}$		19.5	30	m Ω
BATTERY CHARGER						
V_{BATREG_RANGE}	Charge voltage program range		3.856		4.624	V
V_{BATREG_STEP}	Charge voltage step			32		mV
V_{BATREG}	Charge voltage setting	$V_{REG}(\text{REG04}[7:3]) = 4.208\text{ V}$ (01011), V, $-40 \leq T_J \leq 85^{\circ}\text{C}$	4.187	4.208	4.229	V
		$V_{REG}(\text{REG04}[7:3]) = 4.352\text{ V}$ (01111), V, $-40 \leq T_J \leq 85^{\circ}\text{C}$	4.330	4.352	4.374	V
V_{BATREG_ACC}	Charge voltage setting accuracy	$V_{BAT} = 4.208\text{ V}$ or $V_{BAT} = 4.352\text{ V}$, $-40 \leq T_J \leq 85^{\circ}\text{C}$	-0.5%		0.5%	

Electrical Characteristics (continued)

$V_{VAC_UVLOZ} < V_{VAC} < V_{VAC_OV}$ and $V_{VAC} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to 125°C and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
	$I_{CHG_REG_RANGE}$	Charge current regulation range	0		3000	mA	
	$I_{CHG_REG_STEP}$	Charge current regulation step		60		mA	
	I_{CHG_REG}	Charge current regulation setting	$I_{CHG} = 240\text{ mA}$, $V_{VBAT} = 3.1\text{ V}$ or $V_{VBAT} = 3.8\text{ V}$	0.216	0.24	0.264	A
	$I_{CHG_REG_ACC}$	Charge current regulation accuracy	$I_{CHG} = 240\text{ mA}$, $V_{VBAT} = 3.1\text{ V}$ or $V_{VBAT} = 3.8\text{ V}$	-10%		10%	
	I_{CHG_REG}	Charge current regulation setting	$I_{CHG} = 720\text{ mA}$, $V_{VBAT} = 3.1\text{ V}$ or $V_{VBAT} = 3.8\text{ V}$	0.685	0.720	0.755	A
	$I_{CHG_REG_ACC}$	Charge current regulation accuracy	$I_{CHG_REG} = 720\text{ mA}$, $V_{BAT} = 3.1\text{ V}$ or $V_{BAT} = 3.8\text{ V}$	-5.0%		-5.0%	
	I_{CHG_REG}	Charge current regulation setting	$I_{CHG} = 1.38\text{ A}$, $V_{VBAT} = 3.1\text{ V}$ or $V_{VBAT} = 3.8\text{ V}$	1.311	1.380	1.449	A
	$I_{CHG_REG_ACC}$	Charge current regulation accuracy	$I_{CHG} = 720\text{ mA}$ or $I_{CHG} = 1.38\text{ A}$, $V_{VBAT} = 3.1\text{ V}$ or $V_{VBAT} = 3.8\text{ V}$	-5%		5%	
	$V_{BATLOWV_FALL}$	Battery LOWV falling threshold	$I_{CHG} = 240\text{ mA}$	2.7	2.8	2.9	V
	$V_{BATLOWV_RISE}$	Battery LOWV rising threshold	Pre-charge to fast charge	3.0	3.12	3.24	V
	I_{PRECHG}	Precharge current regulation	$IPRECHG[3:0] = '0010' = 180\text{ mA}$	153	171	189	mA
	I_{PRECHG_ACC}	Precharge current regulation accuracy	$IPRECHG[3:0] = '0010' = 180\text{ mA}$	-15%		5%	
	I_{TERM}	Termination current regulation	$I_{CHG} > 780\text{ mA}$, $ITERM[3:0] = '0010' = 180\text{ mA}$, $V_{VBAT} = 4.208\text{ V}$	150	180	216	mA
	I_{TERM_ACC}	Termination current regulation accuracy	$I_{CHG} > 780\text{ mA}$, $ITERM[3:0] = '0010' = 180\text{ mA}$, $V_{VBAT} = 4.208\text{ V}$	-16.7		20	%
	I_{TERM}	Termination current regulation	$I_{CHG} \leq 780\text{ mA}$, $ITERM[3:0] = '0010' = 180\text{ mA}$	162	180	192	mA
	I_{TERM_ACC}	Termination current regulation accuracy	$I_{CHG} \leq 780\text{ mA}$, $ITERM[3:0] = '0010' = 180\text{ mA}$	-10%		10%	
	I_{TERM}	Termination current regulation	$I_{CHG} = 600\text{ mA}$, $ITERM[3:0] = '0000' = 60\text{ mA}$, $V_{VBAT} = 4.208\text{ V}$	45	60	75	mA
	I_{TERM_ACC}	Termination current regulation accuracy	$I_{CHG} = 600\text{ mA}$, $ITERM[3:0] = '0000' = 60\text{ mA}$, $V_{VBAT} = 4.208\text{ V}$	-25%		25%	
	V_{SHORT}	Battery short voltage	V_{VBAT} falling	1.85	2	2.15	V
	V_{SHORTZ}	Battery short voltage	V_{VBAT} rising	2.15	2.25	2.35	V
	I_{SHORT}	Battery short current	$V_{VBAT} < V_{SHORTZ}$	70	90	110	mA
	V_{RECHG}	Recharge Threshold below V_{BAT_REG}	V_{BAT} falling, $REG04[0] = 0$	90	120	150	mV
	V_{RECHG}	Recharge Threshold below V_{BAT_REG}	V_{BAT} falling, $REG04[0] = 1$	200	230	265	mV
	$I_{SYSLOAD}$	System discharge load current	$V_{SYS} = 4.2\text{ V}$		30		mA
INPUT VOLTAGE AND CURRENT REGULATION							
	V_{INDPM}	Input voltage regulation limit	$V_{INDPM} (REG06[3:0] = 0000) = 3.9\text{ V}$	3.78	3.95	4.1	V
	V_{INDPM_ACC}	Input voltage regulation accuracy		-3%		5%	
	V_{INDPM}	Input voltage regulation limit	$V_{INDPM} (REG06[3:0] = 0110) = 4.4\text{ V}$	4.268	4.4	4.532	V
	V_{INDPM_ACC}	Input voltage regulation accuracy		-3%		3%	
	V_{DPM_VBAT}	Input voltage regulation limit tracking VBAT	$V_{INDPM} = 3.9\text{ V}$, $V_{DPM_VBAT_TRACK} = 300\text{ mV}$, $V_{BAT} = 4.0\text{ V}$	4.171	4.3	4.43	V
	$V_{DPM_VBAT_ACC}$	Input voltage regulation accuracy tracking VBAT		-3%		3%	

Electrical Characteristics (continued)

$V_{VAC_UVLOZ} < V_{VAC} < V_{VAC_OV}$ and $V_{VAC} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to 125°C and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
I _{IINDPM}	V _{VBUS} = 5 V, current pulled from SW, I _{IINDPM} (REG[4:0] = 00100) = 500 mA, $-40 \leq T_J \leq 85^{\circ}\text{C}$	450		500	mA	
	V _{VBUS} = 5 V, current pulled from SW, I _{IINDPM} (REG[4:0] = 01000) = 900 mA, $-40 \leq T_J \leq 85^{\circ}\text{C}$	750		900	mA	
	V _{VBUS} = 5 V, current pulled from SW, I _{IINDPM} (REG[4:0] = 01110) = 1.5 A, $-40 \leq T_J \leq 85^{\circ}\text{C}$	1.3		1.5	A	
I _{IN_START}	Input current limit during system start-up sequence		200		mA	
BAT PIN OVERVOLTAGE PROTECTION						
V _{BATOVP_RISE}	Battery overvoltage threshold	V _{BAT} rising, as percentage of V _{BAT_REG}	103	104	105	%
V _{BATOVP_FALL}	Battery overvoltage threshold	V _{BAT} falling, as percentage of V _{BAT_REG}	101	102	103	%
THERMAL REGULATION AND THERMAL SHUTDOWN						
T _{JUNCTION_REG}	Junction Temperature Regulation Threshold	Temperature Increasing, T _{REG} (REG05[1] = 1) = 110°C		110		°C
T _{JUNCTION_REG}	Junction Temperature Regulation Threshold	Temperature Increasing, T _{REG} (REG05[1] = 0) = 90°C		90		°C
T _{SHUT}	Thermal Shutdown Rising Temperature	Temperature Increasing		160		°C
T _{SHUT_HYST}	Thermal Shutdown Hysteresis			30		°C
JEITA Thermistor Comparator (BUCK MODE)						
V _{T1}	T1 (0°C) threshold, Charge suspended T1 below this temperature.	Charger suspends charge. As Percentage to V _{REGN}	72.4%	73.3%	74.2%	
V _{T1}	Falling	As Percentage to V _{REGN}	69%	71.5%	74%	
V _{T2}	T2 (10°C) threshold, Charge back to I _{CHG} /2 and 4.2 V below this temperature	As percentage of V _{REGN}	67.2%	68%	69%	
V _{T2}	Falling	As Percentage to V _{REGN}	66%	66.8%	67.7%	
V _{T3}	T3 (45°C) threshold, charge back to I _{CHG} and 4.05V above this temperature.	Charger suspends charge. As Percentage to V _{REGN}	43.8%	44.7%	45.8%	
V _{T3}	Falling	As Percentage to V _{REGN}	45.1%	45.7%	46.2%	
V _{T5}	T5 (60°C) threshold, charge suspended above this temperature.	As Percentage to V _{REGN}	33.7%	34.2%	35.1%	
V _{T5}	Falling	As Percentage to V _{REGN}	34.5%	35.3%	36.2%	
COLD OR HOT THERMISTOR COMPARATOR (BOOST MODE)						
V _{BCOLD}	Cold Temperature Threshold, TS pin Voltage Rising Threshold	As Percentage to V _{REGN} (Approx. -20°C w/ 103AT), T _J = -20°C - 125°C	79.5%	80%	80.5%	
V _{BCOLD}	Falling	T _J = -20°C - 125°C	78.5%	79%	79.5%	
V _{BHOT}	Hot Temperature Threshold, TS pin Voltage falling Threshold	As Percentage to V _{REGN} (Approx. 60°C w/ 103AT), T _J = -20°C - 125°C	30.2%	31.2%	32.2%	
V _{BHOT}	Rising	T _J = -20°C - 125°C	33.8%	34.4%	34.9%	

Electrical Characteristics (continued)

$V_{VAC_UVLOZ} < V_{VAC} < V_{VAC_OV}$ and $V_{VAC} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to 125°C and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

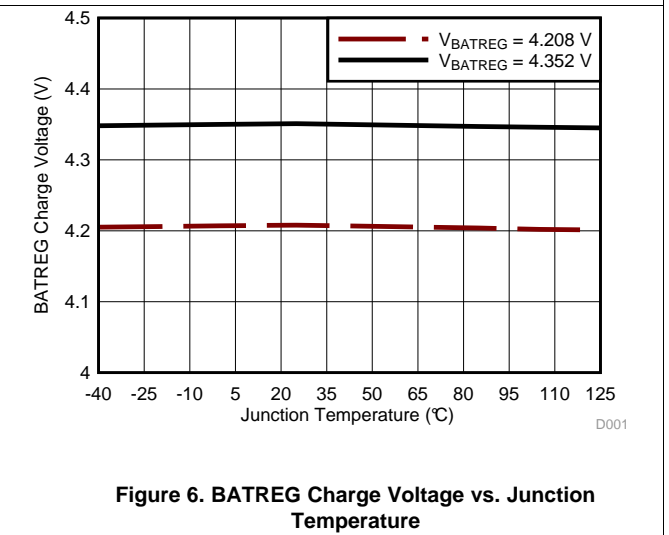
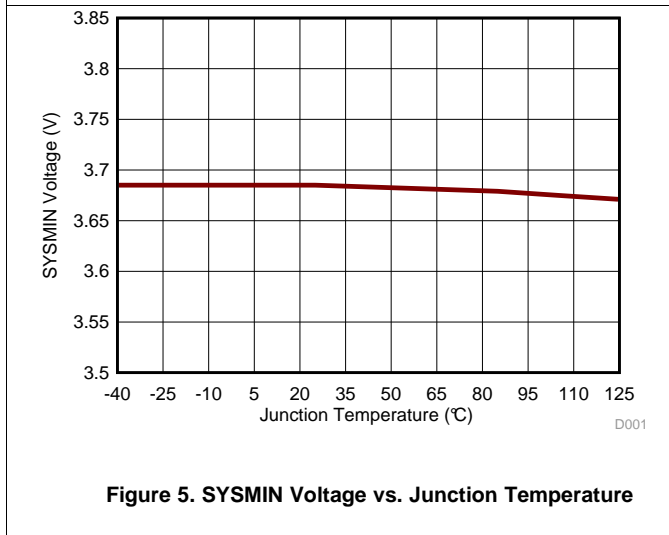
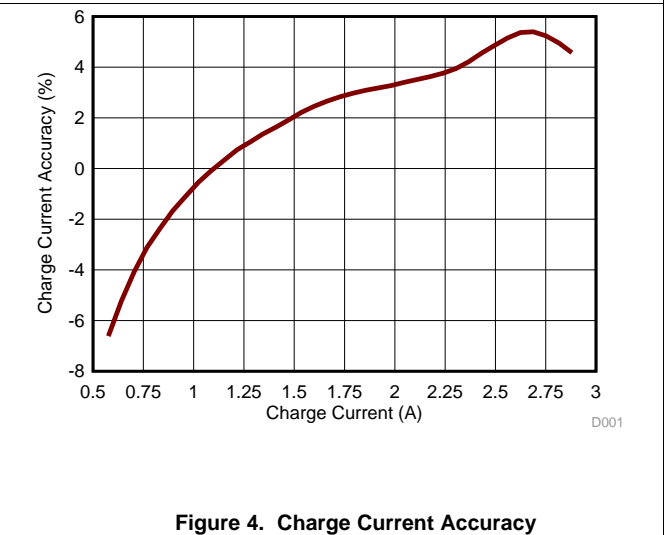
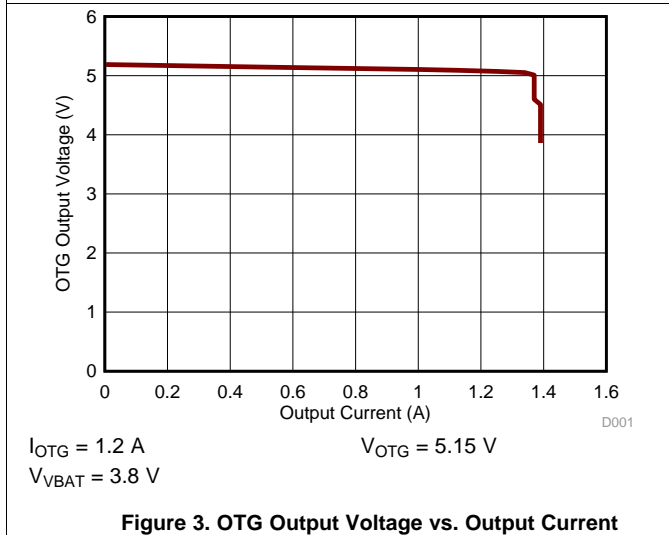
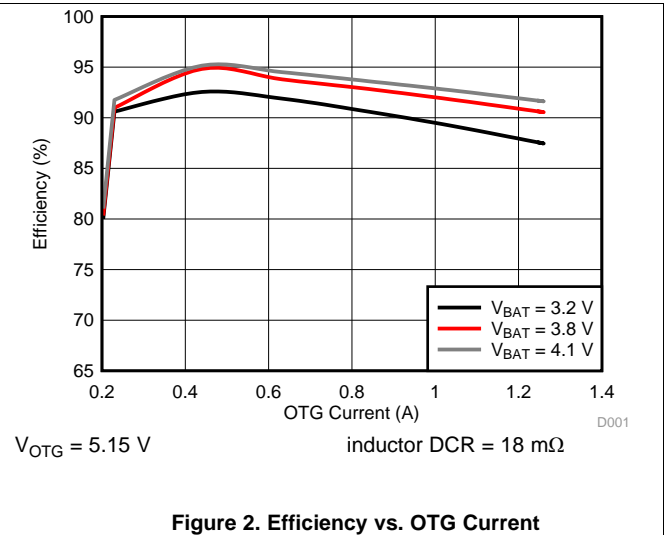
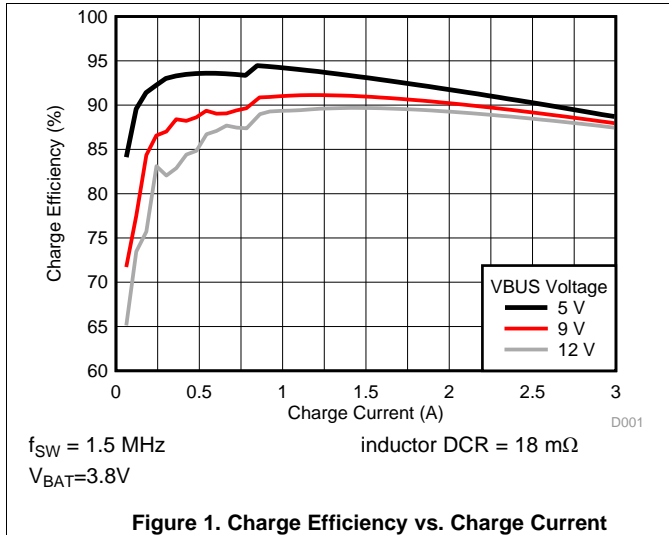
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
CHARGE OVERCURRENT COMPARATOR (CYCLE-BY-CYCLE)						
I_{HSFET_OCP}	HSFET cycle-by-cycle over-current threshold		5.2		8.0	A
I_{BATFET_OCP}	System over load threshold		6.0			A
CHARGE UNDER-CURRENT COMPARATOR (CYCLE-BY-CYCLE)						
V_{LSFET_UCP}	LSFET under-current falling threshold	From sync mode to non-sync mode			160	mA
PWM						
f_{SW}	PWM switching frequency	Oscillator frequency, buck mode	1320	1500	1680	kHz
		Oscillator frequency, boost mode	1150	1412	1660	kHz
D_{MAX}	Maximum PWM duty cycle ⁽¹⁾			97%		
BOOST MODE OPERATION						
V_{OTG_REG}	Boost mode regulation voltage	$V_{VBAT} = 3.8\text{ V}$, $I_{(PMID)} = 0\text{ A}$, $BOOSTV[1:0] = '10' = 5.15\text{ V}$	4.972	5.126	5.280	V
$V_{OTG_REG_ACC}$	Boost mode regulation voltage accuracy	$V_{VBAT} = 3.8\text{ V}$, $I_{(PMID)} = 0\text{ A}$, $BOOSTV[1:0] = '10' = 5.15\text{ V}$	-3		3	%
$V_{BATLOWV_OTG}$	Battery voltage exiting boost mode	V_{VBAT} falling, MIN_VBAT_SEL (REG01[0]) = 0	2.6	2.8	2.9	V
		V_{VBAT} rising, MIN_VBAT_SEL (REG01[0]) = 0	2.9	3.0	3.15	V
		V_{VBAT} falling, MIN_VBAT_SEL (REG01[0]) = 1	2.4	2.5	2.6	V
		V_{VBAT} rising, MIN_VBAT_SEL (REG01[0]) = 1	2.7	2.8	2.9	V
I_{OTG}	OTG mode output current	$BOOST_LIM$ (REG02[7]) = 1	1.2	1.4	1.6	A
$I_{OTG_OCP_ACC}$	Boost mode RBFET over-current protection accuracy	$BOOST_LIM = 0.5\text{ A}$ (REG02[7] = 0)	0.5		0.722	A
V_{OTG_OVP}	OTG overvoltage threshold	Rising threshold	5.55	5.8	6.15	V
I_{OTG_HSZCP}	HSFET under current falling threshold			100		mA
REGN LDO						
V_{REGN}	REGN LDO output voltage	$V_{VBUS} = 9\text{ V}$, $I_{REGN} = 40\text{ mA}$	5.6	6	6.55	V
V_{REGN}	REGN LDO output voltage	$V_{VBUS} = 5\text{ V}$, $I_{REGN} = 20\text{ mA}$	4.6	4.7	4.8	V
LOGIC I/O PIN CHARACTERISTICS (\overline{CE}, PSEL, SCL, SDA,, \overline{INT})						
V_{ILO}	Input low threshold \overline{CE}				0.4	V
V_{IH}	Input high threshold \overline{CE}		1.3			V
I_{BIAS}	High-level leakage current \overline{CE}	Pull up rail 1.8 V			1	μA
V_{ILO}	Input low threshold PSEL				0.4	V
V_{IH}	Input high threshold PSEL		1.3			V
I_{BIAS}	High-level leakage current PSEL	Pull up rail 1.8V			1	μA
LOGIC I/O PIN CHARACTERISTICS (\overline{PG}, STAT)						
V_{OL}	Low-level output voltage				0.4	V

(1) Specified by design. Not production tested.

7.5 Timing Requirements

Parameter		Additional Information	MIN	NOM	MAX	UNIT
VBUS/BAT POWER UP						
t _{ACOV}	VAC OVP reaction time	VAC rising above ACOV threshold to turn off Q2		200		ns
t _{BADSRC}	Bad adapter detection duration			30		ms
BATTERY CHARGER						
t _{TERM_DGL}	Deglintch time for charge termination			250		ms
t _{RECHG_DGL}	Deglintch time for recharge			250		ms
t _{SYSOVLD_DGL}	System over-current deglintch time to turn off Q4			100		µs
t _{BATOVP}	Battery over-voltage deglintch time to disable charge			1		µs
t _{SAFETY}	Typical Charge Safety Timer Range	CHG_TIMER = 1	8	10	12	hr
t _{TOP_OFF}	Typical Top-Off Timer Range	TOP_OFF_TIMER[1:0] = 10 (30 min)	24	30	36	min
QON TIMING						
t _{SHIPMODE}	/QON low time to turn on BATFET and exit ship mode	-10°C ≤ T _J ≤ 60°C	0.9		1.3	s
t _{QON_RST_2}	\overline{QON} low time to reset BATFET	-10°C ≤ T _J ≤ 60°C	8		12	s
t _{BATFET_RST}	BATFET off time during full system reset	-10°C ≤ T _J ≤ 60°C	250		400	ms
t _{SM_DLY}	Enter ship mode delay	-10°C ≤ T _J ≤ 60°C	10		15	s
DIGITAL CLOCK AND WATCHDOG TIMER						
t _{WDT}	REG05[4]=1	REGN LDO disabled		40		s
f _{LPDIG}	Digital Low Power Clock	REGN LDO disabled		30		kHz
f _{DIG}	Digital Clock	REGN LDO enabled		500		kHz
f _{SCL}	SCL clock frequency				400	kHz

7.6 Typical Characteristics



Typical Characteristics (continued)

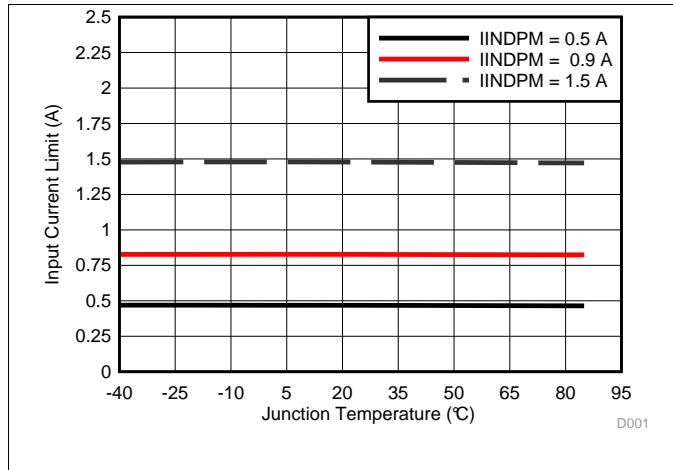


Figure 7. Input Current Limit vs. Junction Temperature

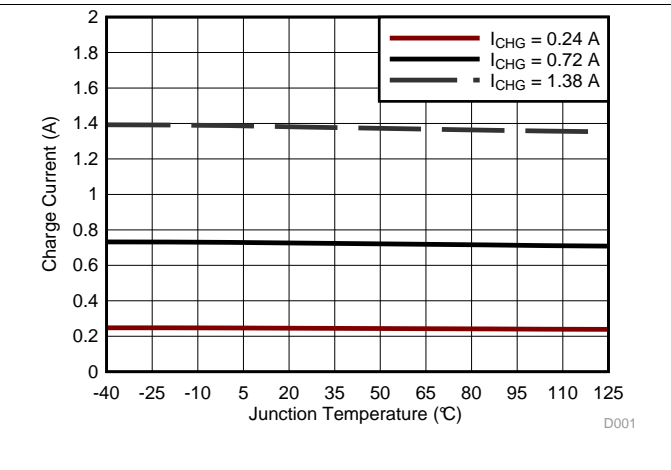


Figure 8. Charge Current vs. Junction Temperature

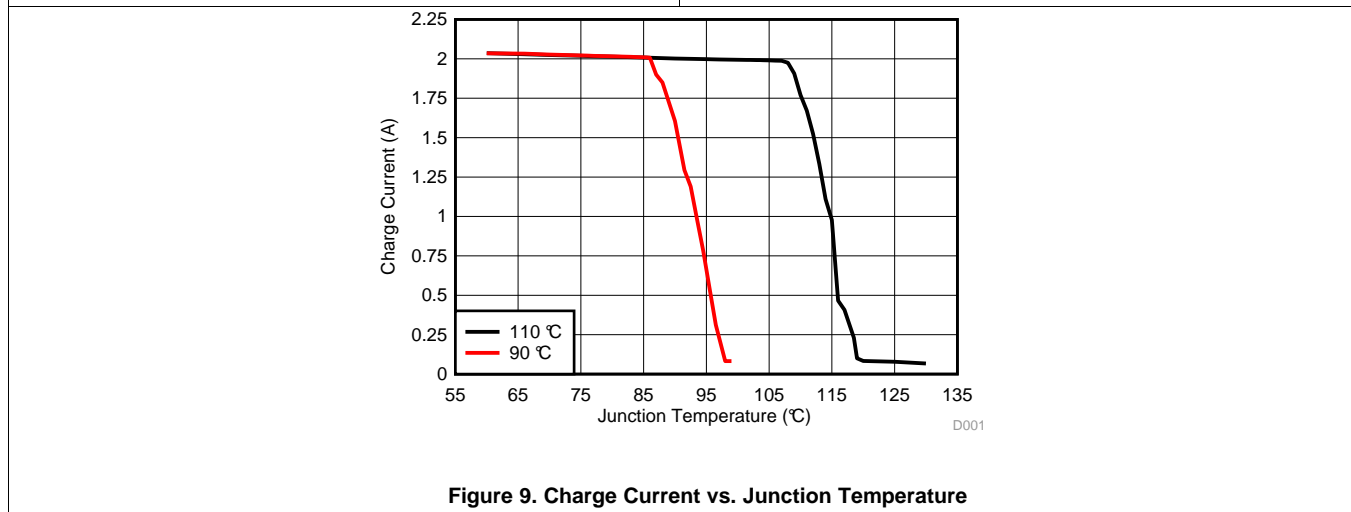


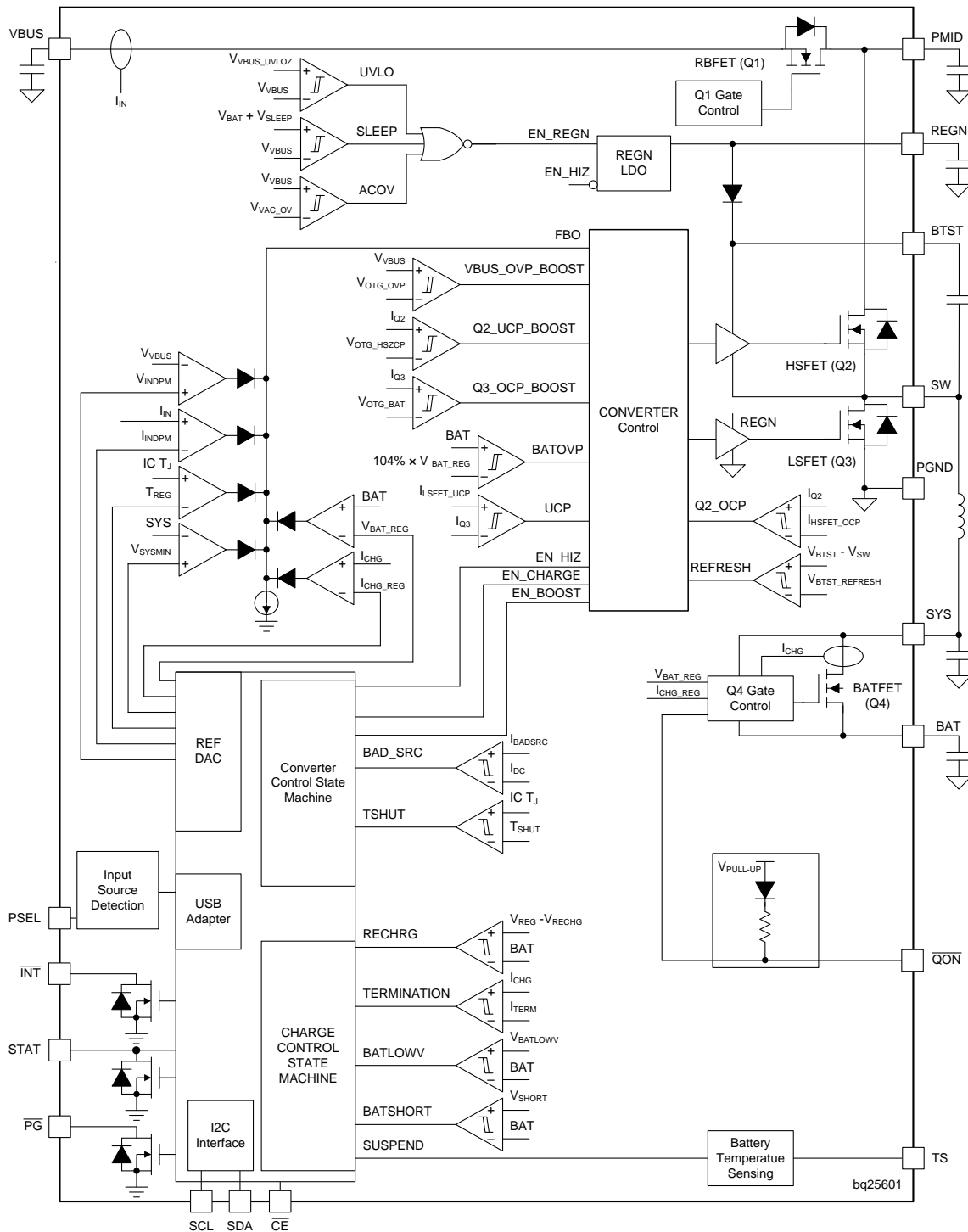
Figure 9. Charge Current vs. Junction Temperature

8 Detailed Description

8.1 Overview

The bq25601 device is a highly integrated 3.0-A switch-mode battery charger for single cell Li-Ion and Li-polymer battery. It includes the input reverse-blocking FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET, Q4), and bootstrap diode for the high-side gate drive.

8.2 Functional Block Diagram



8.3 Feature Description

The bq25601 is an I²C controlled power path management device and a single cell Li-Ion battery charger. It integrates the input reverse-blocking FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET, Q4) between system and battery. The device also integrates the bootstrap diode for the high-side gate drive.

8.4 Device Functional Modes

8.4.1 Power-On-Reset (POR)

The device powers internal bias circuits from the higher voltage of VBUS and BAT. When VBUS rises above V_{VBUS_UVLOZ} or BAT rises above V_{BAT_UVLOZ} , the sleep comparator, battery depletion comparator and BATFET driver are active. I²C interface is ready for communication and all the registers are reset to default value. The host can access all the registers after POR.

8.4.2 Device Power Up from Battery without input Source

If only battery is present and the voltage is above depletion threshold ($V_{BAT_DPL_RISE}$), the BATFET turns on and connects battery to system. The REGN stays off to minimize the quiescent current. The low RDSON of BATFET and the low quiescent current on BAT minimize the conduction loss and maximize the battery run time.

The device always monitors the discharge current through BATFET (*Supplement Mode*). When the system is overloaded or shorted ($I_{BAT} > I_{BATFET_OCP}$), the device turns off BATFET immediately and set BATFET_DIS bit to indicate BATFET is disabled until the input source plugs in again or one of the methods described in *BATFET Enable (Exit Shipping Mode)* is applied to re-enable BATFET.

8.4.3 Power Up from input Source

When an input source is plugged in, the device checks the input source voltage to turn on REGN LDO and all the bias circuits. It detects and sets the input current limit before the buck converter is started. The power up sequence from input source is as listed:

1. Power Up REGN LDO
2. Poor Source Qualification
3. *input Source Type Detection* based on PSEL to set default input Current Limit (IINDPM) register or input source type
4. input Voltage Limit Threshold Setting (VINDPM threshold)
5. Converter Power-up

8.4.3.1 Power Up REGN Regulation

The REGN LDO supplies internal bias circuits as well as the HSFET and LSFET gate drive. The LDO also provides bias rail to TS external resistors. The pull-up rail of STAT can be connected to REGN as well. The REGN is enabled when all the below conditions are valid:

1. V_{VAC} above $V_{VAC_PRESENT}$
2. V_{VAC} above $V_{BAT} + V_{SLEEPZ}$ in buck mode or V_{BUS} below $V_{BAT} + V_{SLEEP}$ in boost mode
3. After 220-ms delay is completed

If one of the above conditions is not valid, the device is in high impedance mode (HIZ) with REGN LDO off. The device draws less than I_{VBUS_HIZ} from VBUS during HIZ state. The battery powers up the system when the device is in HIZ.

8.4.3.2 Poor Source Qualification

After REGN LDO powers up, the device checks the current capability of the input source. The input source has to meet the following requirements in order to start the buck converter.

1. VBUS voltage below V_{VAC_OV}
2. VBUS voltage above $V_{VBUSMIN}$ when pulling I_{BADSRC} (typical 30 mA)

Once the input source passes all the conditions above, the status register bit VBUS_GD is set high and the \overline{INT} pin is pulsed to signal to the host. If the device fails the poor source detection, it repeats poor source qualification every 2 seconds.

8.4.3.3 input Source Type Detection

After the VBUS_GD bit is set and REGN LDO is powered, the charger device runs input source detection through the PSEL pin. The bq25601 sets input current limit through PSEL pin.

Device Functional Modes (continued)

After input source type detection is completed, an INT pulse is asserted to the host. In addition, the following registers and pin are changed:

1. input Current Limit (IINDPM) register is changed to set current limit
2. PG_STAT bit is set
3. VBUS_STAT bit is updated to indicate USB or other input source

The host can over-write IINDPM register to change the input current limit if needed. The charger input current is always limited by the IINDPM register.

8.4.3.3.1 PSEL pins sets input current limit in bq25601

The bq25601 has PSEL pin for input current limit setting to interface with USB PHY. It directly takes the USB PHY device output to decide whether the input is USB host or charging port. When the device operates in host-control mode, the host needs to IINDET_EN bit to read the PSEL value and update the IINDPM register. When the device is in default mode, PSEL value updates IINDPM in real time.

Table 1. input Current Limit Setting from PSEL

input Detection	PSEL Pin	inPUT CURRENT LIMIT (ILIM)	VBUS_STAT
USB SDP	High	500 mA	001
Adapter	Low	2.4A	011

8.4.3.4 input Voltage Limit Threshold Setting (VINDPM Threshold)

The device supports wide range of input voltage limit (3.9 V – 5.4V) for USB. The default VINDPM setting is 4.5V.

In addition to absolute VINDPM voltage setting, the device supports dynamic VINDPM settings which tracks the battery voltage. This function can be enabled via the VDPM_BAT_TRACK[1:0] register bits. When enabled, the actual input voltage limit will be the higher of the VINDPM register and VBAT + VDPM_BAT_TRACK offset.

8.4.3.5 Converter Power-Up

After the input current limit is set, the converter is enabled and the HSFET and LSFET start switching. If battery charging is disabled, BATFET turns off. Otherwise, BATFET stays on to charge the battery.

The device provides soft-start when system rail is ramped up. When the system rail is below 2.2 V, the input current limit is set to the lower of 200 mA or IINDPM register setting. After the system rises above 2.2 V, the device limits input current to the value of IINDPM register.

As a battery charger, the device deploys a highly efficient 1.5 MHz step-down switching regulator. The fixed frequency oscillator keeps tight control of the switching frequency under all conditions of input voltage, battery voltage, charge current and temperature, simplifying output filter design.

A type III compensation network allows using ceramic capacitors at the output of the converter. An internal sawtooth ramp is compared to the internal error control signal to vary the duty cycle of the converter. The ramp height is proportional to the PMID voltage to cancel out any loop gain variation due to a change in input voltage.

In order to improve light-load efficiency, the device switches to PFM control at light load when battery is below minimum system voltage setting or charging is disabled. During the PFM operation, the switching duty cycle is set by the voltage ratio of SYS and VBUS. The PFM_DIS bit can be used to prevent PFM operation in either buck or boost configuration.

8.4.4 Boost Mode Operation From Battery

The device supports boost converter operation to deliver power from the battery to other portable devices through USB port. The boost mode output current rating meets the USB On-The-Go 500 mA output requirement. The maximum output current is up to 1.2 A. The boost operation can be enabled if the conditions are valid:

1. BAT above V_{OTG_BAT}
2. VBUS less than $BAT + V_{SLEEP}$ (in sleep mode)
3. Boost mode operation is enabled (OTG_CONFIG bit = 1)
4. Voltage at TS (thermistor) pin is within acceptable range ($V_{BHOT} < V_{TS} < V_{BCOLD}$)
5. After 30-ms delay from boost mode enable

in boost mode, the device employs 1.4 MHz step-up switching regulator.

During boost mode, the status register VBUS_STAT bits is set to 111, the VBUS output is 5.15 V by default and the output current can reach up to 1.2 A, selected through I²C (BOOST_LIM bit). The boost output is maintained when BAT is above V_{OTG_BAT} threshold.

When OTG is enabled, the device starts up with PFM and later transits to PWM to minimize the overshoot. The PFM_DIS bit can be used to prevent PFM operation in either buck or boost configuration.

8.4.5 Host Mode and Default Mode

8.4.5.1 Host Mode and Default Mode in bq25601

The bq25601 is a host controlled charger, but it can operate in default mode without host management. In default mode, the device can be used as an autonomous charger with no host or while host is in sleep mode. When the charger is in default mode, WATCHDOG_FAULT bit is HIGH. When the charger is in host mode, WATCHDOG_FAULT bit is LOW.

After power-on-reset, the device starts in default mode with watchdog timer expired, or default mode. All the registers are in the default settings. During default mode, any change on PSEL pin will make real time IINDPM register changes.

In default mode, the device keeps charging the battery with default 10-hour fast charging safety timer. At the end of the 10-hour, the charging is stopped and the buck converter continues to operate to supply system load.

Writing a 1 to the WD_RST bit transitions the charger from default mode to host mode. All the device parameters can be programmed by the host. To keep the device in host mode, the host has to reset the watchdog timer by writing 1 to WD_RST bit before the watchdog timer expires (WATCHDOG_FAULT bit is set), or disable watchdog timer by setting WATCHDOG bits = 00.

When the watchdog timer expires (WATCHDOG_FAULT bit = 1), the device returns to default mode and all registers are reset to default values except IINDPM, VINDPM, BATFET_RST_EN, BATFET_DLY, and BATFET_DIS bits.

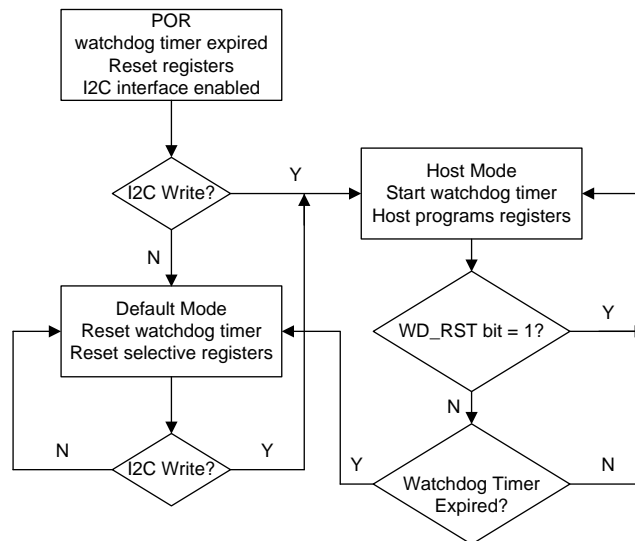


Figure 10. Watchdog Timer Flow Chart

8.4.6 Power Path Management

The device accommodates a wide range of input sources from USB, wall adapter, to car battery. The device provides automatic power path selection to supply the system (SYS) from input source (VBUS), battery (BAT), or both.

8.4.7 Battery Charging management

The device charges 1-cell Li-Ion battery with up to 3.0-A charge current for high capacity tablet battery. The 19.5-mΩ BATFET improves charging efficiency and minimize the voltage drop during discharging.

8.4.7.1 Autonomous Charging Cycle

With battery charging is enabled (CHG_CONFIG bit = 1 and \overline{CE} pin is LOW), the device autonomously completes a charging cycle without host involvement. The device default charging parameters are listed in [Table 2](#). The host can always control the charging operations and optimize the charging parameters by writing to the corresponding registers through I²C.

Table 2. Charging Parameter Default Setting

Default Mode	bq25601
Charging voltage	4.208V
Charging current	2.048 A
Pre-charge current	180 mA
Termination current	180 mA
Temperature profile	JEITA
Safety timer	10 hours

A new charge cycle starts when the following conditions are valid:

- Converter starts
- Battery charging is enabled (CHG_CONFIG bit = 1 and I_{CHG} register is not 0 mA and \overline{CE} is low)
- No thermistor fault on TS
- No safety timer fault
- BATFET is not forced to turn off (BATFET_DIS bit = 0)

The charger device automatically terminates the charging cycle when the charging current is below termination threshold, charge voltage is above recharge threshold, and device not is in DPM mode or thermal regulation.

When a full battery voltage is discharged below recharge threshold (selectable through VRECHG bit), the device automatically starts a new charging cycle. After the charge is done, toggle \overline{CE} pin or CHG_CONFIG bit can initiate a new charging cycle.

The STAT output indicates the charging status: charging (LOW), charging complete or charge disable (HIGH) or charging fault (Blinking). The STAT output can be disabled by setting EN_ICHG_MON bits = 11. In addition, the status register (CHRG_STAT) indicates the different charging phases: 00-charging disable, 01-precharge, 10-fast charge (constant current) and constant voltage mode, 11-charging done. Once a charging cycle is completed, an INT is asserted to notify the host.

8.4.7.2 Battery Charging Profile

The device charges the battery in five phases: battery short, preconditioning, constant current, constant voltage and top-off trickle charging (optional). At the beginning of a charging cycle, the device checks the battery voltage and regulates current and voltage accordingly.

Table 3. Charging Current Setting

V _{BAT}	CHARGING CURRENT	REGISTER DEFAULT SETTING	CHRG_STAT
< 2.2 V	I _{SHORT}	100 mA	01
2.2 V to 3 V	I _{PRECHG}	180 mA	01
> 3 V	I _{CHG}	2.048 A	10

If the charger device is in DPM regulation or thermal regulation during charging, the actual charging current will be less than the programmed value. In this case, termination is temporarily disabled and the charging safety timer is counted at half the clock rate.

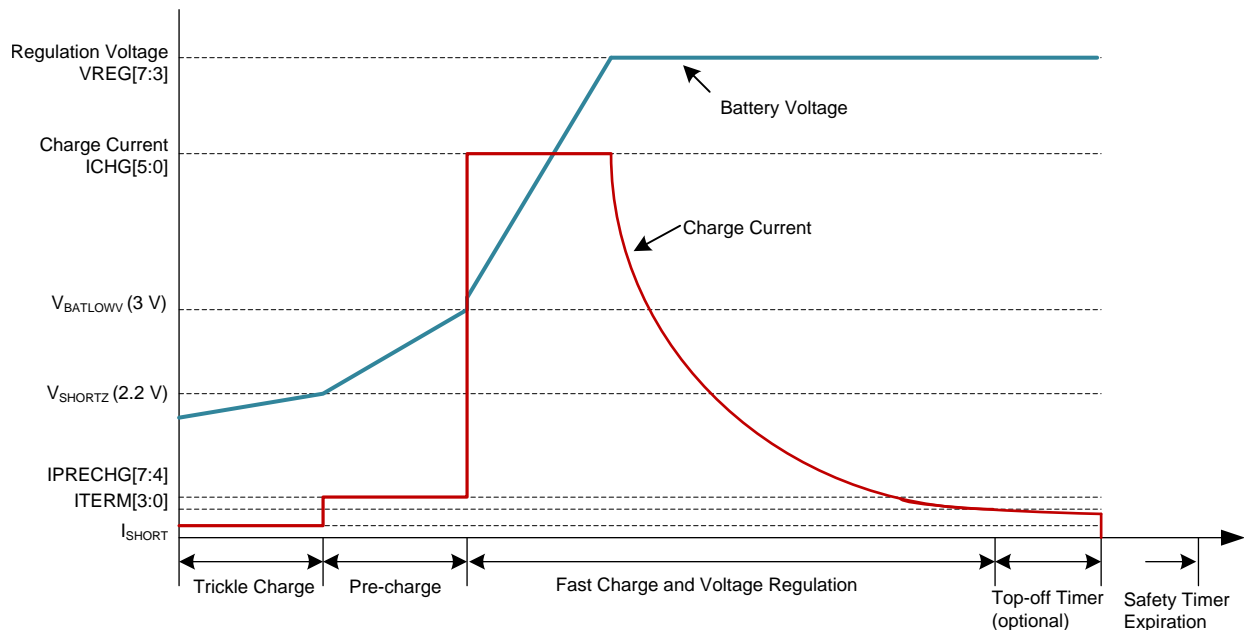


Figure 11. Battery Charging Profile

8.4.7.3 Charging Termination

The device terminates a charge cycle when the battery voltage is above recharge threshold, and the current is below termination current. After the charging cycle is completed, the BATFET turns off. The converter keeps running to power the system, and BATFET can turn on again to engage Supplement Mode.

When termination occurs, the status register CHRG_STAT is set to 11, and an INT pulse is asserted to the host. Termination is temporarily disabled when the charger device is in input current, voltage or thermal regulation. Termination can be disabled by writing 0 to EN_TERM bit prior to charge termination.

At low termination currents (25 mA-50 mA), due to the comparator offset, the actual termination current may be 10 mA-20 mA higher than the termination target. In order to compensate for comparator offset, a programmable top-off timer can be applied after termination is detected. The termination timer will follow safety timer constraints, such that if safety timer is suspended, so will the termination timer. Similarly, if safety timer is doubled, so will the termination timer. TOPOFF_ACTIVE bit reports whether the top off timer is active or not. The host can read CHRГ_STAT and TOPOFF_ACTIVE to find out the termination status.

Top off timer gets reset at one of the following conditions:

1. Charge disable to enable
2. Termination status low to high
3. REG_RST register bit is set

The top-off timer settings are read in once termination is detected by the charger. Programming a top-off timer value after termination will have no effect unless a recharge cycle is initiated. An INT is asserted to the host when entering top-off timer segment as well as when top-off timer expires.

8.4.7.4 Thermistor Qualification

The charger device provides a single thermistor input for battery temperature monitor.

8.4.7.5 JEITA Guideline Compliance During Charging Mode

To improve the safety of charging Li-ion batteries, JEITA guideline was released on April 20, 2007. The guideline emphasized the importance of avoiding a high charge current and high charge voltage at certain low and high temperature ranges.

To initiate a charge cycle, the voltage on TS pin must be within the VT1 to VT5 thresholds. If TS voltage exceeds the T1-T5 range, the controller suspends charging and waits until the battery temperature is within the T1 to T5 range.

At cool temperature (T1-T2), JEITA recommends the charge current to be reduced to half of the charge current or lower. At warm temperature (T3-T5), JEITA recommends charge voltage less than 4.1 V.

The charger provides flexible voltage/current settings beyond the JEITA requirement. The voltage setting at warm temperature (T3-T5) can be VREG or 4.1V (configured by JEITA_VSET). The current setting at cool temperature (T1-T2) can be further reduced to 20% of fast charge current (JEITA_ISET).

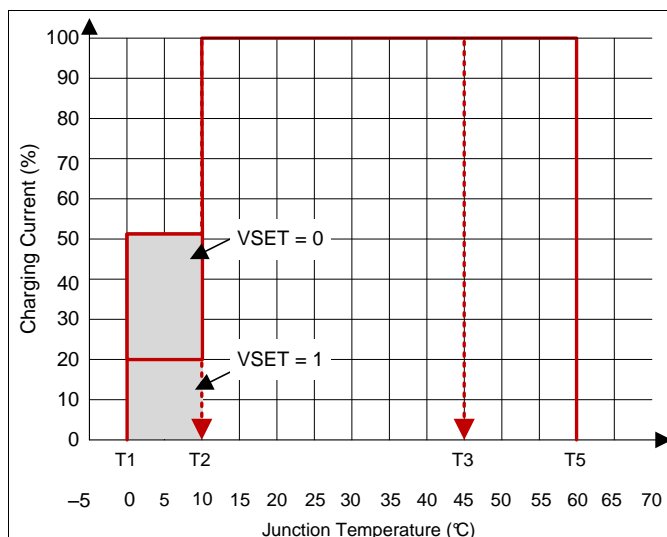


Figure 12. JEITA Profile: Charging Current

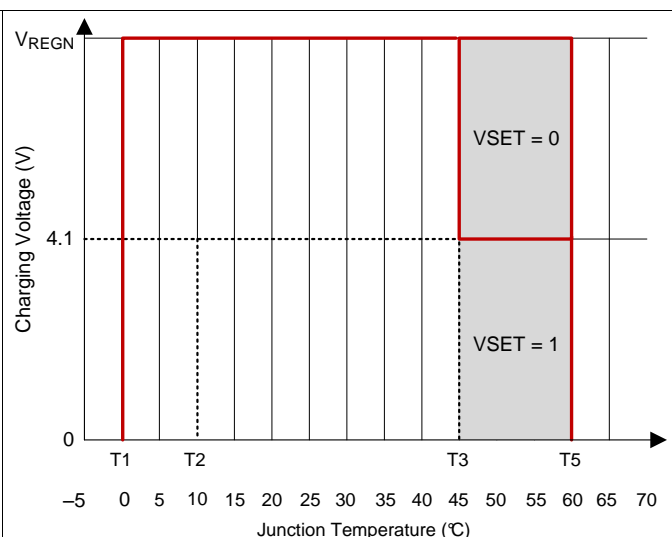


Figure 13. JEITA Profile: Charging Voltage

The resistor bias network has been updated as below.

$$RT2 = \frac{V_{REGN} \times RTH_{COLD} \times RTH_{HOT} \times \left(\frac{1}{VT1} - \frac{1}{VT5} \right)}{RTH_{HOT} \times \left(\frac{V_{REGN}}{VT5} - 1 \right) - RTH_{COLD} \times \left(\frac{V_{REGN}}{VT1} - 1 \right)} \quad (1)$$

$$RT1 = \frac{\left(\left(\frac{V_{REGN}}{VT1} \right) - 1 \right)}{\left(\frac{1}{RT2} \right) + \left(\frac{1}{RTH_{COLD}} \right)} \quad (2)$$

Select 0°C to 60°C range for Li-ion or Li-polymer battery, (Actual example is TBD based on threshold design)

- $RHT_{COLD} = 27.28 \text{ K}\Omega$
- $RHT_{HOT} = 3.02 \text{ K}\Omega$
- $RT1 = 5.23 \text{ K}\Omega$
- $RT2 = 30.9 \text{ K}\Omega$

8.4.7.6 Boost Mode Thermistor Monitor during Battery Discharge Mode

For battery protection during boost mode, the device monitors the battery temperature to be within the VBCOLDx to VBHOTx thresholds unless boost mode temperature is disabled by setting BHOT bits to 11. When temperature is outside of the temperature thresholds, the boost mode is suspended. In addition, VBUS_STAT bits are set to 000 and NTC_FAULT is reported. Once temperature returns within thresholds, the boost mode is recovered and NTC_FAULT is cleared.

When BHOT bits are set to 11, the temperature monitor reports the NTC_FAULT, but the boost mode operation is not suspended.

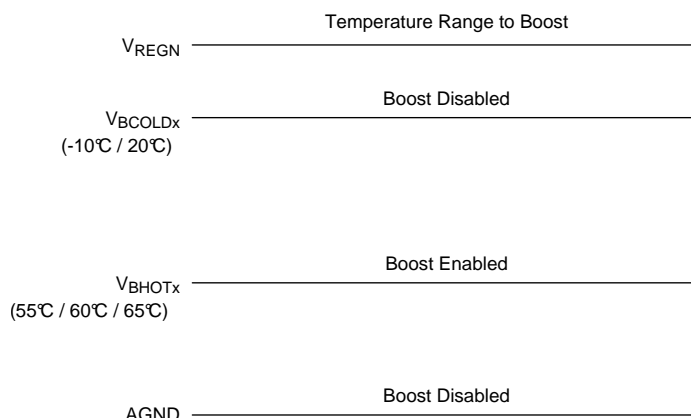


Figure 14. TS Pin Thermistor Sense Threshold in Boost Mode

8.4.7.7 Charging Safety Timer

The device has built-in safety timer to prevent extended charging cycle due to abnormal battery conditions. The safety timer is 2 hours when the battery is below V_{BAT_LOWV} threshold. The user can program fast charge safety timer through I²C (CHG_TIMER bits). When safety timer expires, the fault register CHRG_FAULT bits are set to 11 and an INT is asserted to the host. The safety timer feature can be disabled through I²C by setting EN_TIMER bit.

During input voltage, current, JEITA cool or thermal regulation, the safety timer counts at half clock rate as the actual charge current is likely to be below the register setting. For example, if the charger is in input current regulation (IDPM_STAT = 1) throughout the whole charging cycle, and the safety time is set to 5 hours, the safety timer will expire in 10 hours. This half clock rate feature can be disabled by writing 0 to TMR2X_EN bit.

During the fault, timer is suspended. Once the fault goes away, fault resumes. If user stops the current charging cycle, and start again, timer gets reset (toggle CE pin or CHRG_CONFIG bit).

8.4.7.8 Narrow VDC Architecture

The device deploys Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by SYS_Min bits. Even with a fully depleted battery, the system is regulated above the minimum system voltage (default 3.5 V).

When the battery is below minimum system voltage setting, the BATFET operates in linear mode (LDO mode), and the system is typically 180 mV above the minimum system voltage setting. As the battery voltage rises above the minimum system voltage, BATFET is fully on and the voltage difference between the system and battery is the VDS of BATFET.

When the battery charging is disabled and above minimum system voltage setting or charging is terminated, the system is always regulated at typically 50mV above battery voltage. The status register VSYS_STAT bit goes high when the system is in minimum system voltage regulation.

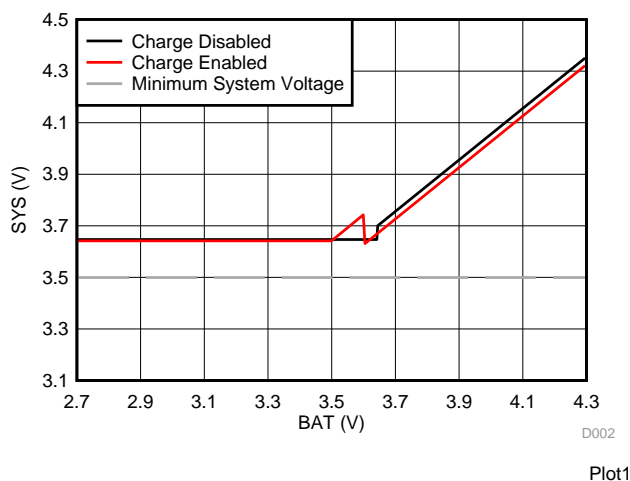


Figure 15. System Voltage vs Battery Voltage

8.4.7.9 Dynamic Power management

To meet maximum current limit in USB spec and avoid over loading the adapter, the device features Dynamic Power management (DPM), which continuously monitors the input current and input voltage. When input source is over-loaded, either the current exceeds the input current limit (IIDPM) or the voltage falls below the input voltage limit (VINDPM). The device then reduces the charge current until the input current falls below the input current limit and the input voltage rises above the input voltage limit.

When the charge current is reduced to zero, but the input source is still overloaded, the system voltage starts to drop. Once the system voltage falls below the battery voltage, the device automatically enters the supplement mode where the BATFET turns on and battery starts discharging so that the system is supported from both the input source and battery.

During DPM mode, the status register bits VDPM_STAT (VINDPM) or IDPM_STAT (IIDPM) goes high. [Figure 16](#) shows the DPM response with 9-V/1.2-A adapter, 3.2-V battery, 2.8-A charge current and 3.4-V minimum system voltage setting.

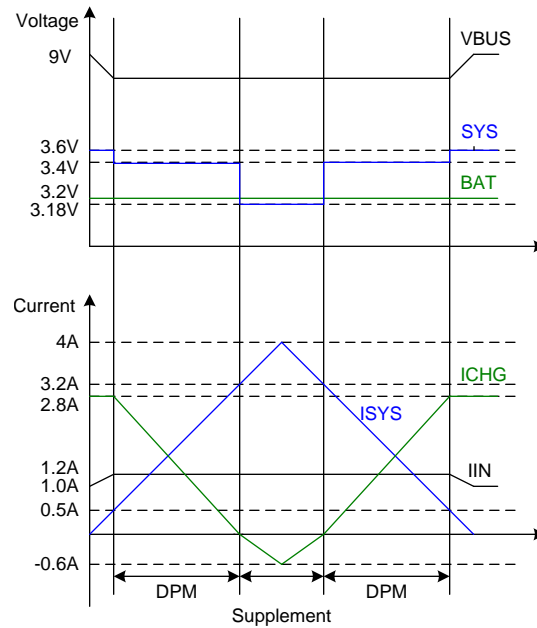


Figure 16. DPM Response

8.4.7.10 Supplement Mode

When the system voltage falls 180 mV ($V_{BAT} > V_{SYSMin}$) or 45 mV ($V_{BAT} < V_{SYSMin}$) below the battery voltage, the BATFET turns on and the BATFET gate is regulated the gate drive of BATFET so that the minimum BATFET VDS stays at 30 mV when the current is low. This prevents oscillation from entering and exiting the supplement mode.

As the discharge current increases, the BATFET gate is regulated with a higher voltage to reduce RDSON until the BATFET is in full conduction. At this point onwards, the BATFET VDS linearly increases with discharge current. Figure 17 shows the V-I curve of the BATFET gate regulation operation. BATFET turns off to exit supplement mode when the battery is below battery depletion threshold.

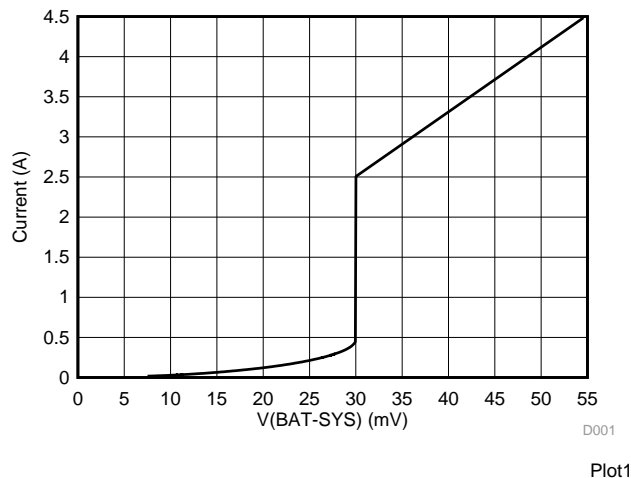


Figure 17. BATFET V-I Curve

8.4.8 Shipping Mode and $\overline{\text{QON}}$ Pin

8.4.8.1 BATFET Disable Mode (Shipping Mode)

To extend battery life and minimize power when system is powered off during system idle, shipping, or storage, the device can turn off BATFET so that the system voltage is zero to minimize the battery leakage current. When the host set BATFET_DIS bit, the charger can turn off BATFET immediately or delay by $t_{\text{SM_DLY}}$ as configured by BATFET_DLY bit.

8.4.8.2 BATFET Enable (Exit Shipping Mode)

When the BATFET is disabled (in shipping mode) and indicated by setting BATFET_DIS, one of the following events can enable BATFET to restore system power:

1. Plug in adapter
2. Clear BATFET_DIS bit
3. Set REG_RST bit to reset all registers including BATFET_DIS bit to default (0)
4. A logic high to low transition on $\overline{\text{QON}}$ pin with t_{SHIPMODE} deglitch time to enable BATFET to exit shipping mode

8.4.8.3 BATFET Full System Reset

The BATFET functions as a load switch between battery and system when input source is not plugged-in. By changing the state of BATFET from on to off, systems connected to SYS can be effectively forced to have a power-on-reset. The $\overline{\text{QON}}$ pin supports push-button interface to reset system power without host by changing the state of BATFET.

When the $\overline{\text{QON}}$ pin is driven to logic low for $t_{\text{QON_RST}}$ while input source is not plugged in and BATFET is enabled (BATFET_DIS = 0), the BATFET is turned off for $t_{\text{BATFET_RST}}$ and then it is re-enabled to reset system power. This function can be disabled by setting BATFET_RST_EN bit to 0.

8.4.8.4 $\overline{\text{QON}}$ Pin Operations

The $\overline{\text{QON}}$ pin incorporates two functions to control BATFET.

1. BATFET Enable: A $\overline{\text{QON}}$ logic transition from high to low with longer than t_{SHIPMODE} deglitch turns on BATFET and exit shipping mode
2. BATFET Reset: When $\overline{\text{QON}}$ is driven to logic low by at least $t_{\text{QON_RST}}$ while adapter is not plugged in (and BATFET_DIS = 0), the BATFET is turned off for $t_{\text{BATFET_RST}}$. The BATFET is re-enabled after $t_{\text{BATFET_RST}}$ duration. This function allows systems connected to SYS to have power-on-reset. This function can be disabled by setting BATFET_RST_EN bit to 0.

Figure 18 shows the sample external configurations for each.

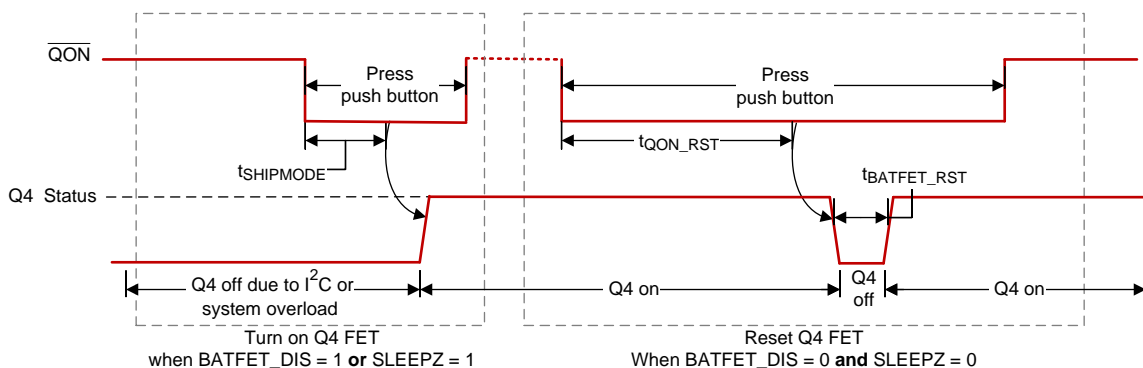


Figure 18. $\overline{\text{QON}}$ Timing

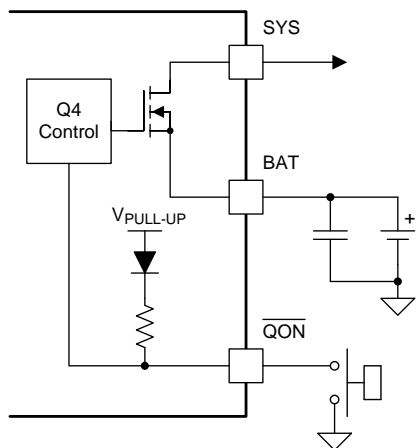


Figure 19. $\overline{\text{QON}}$ Circuit

8.4.9 Status Outputs ($\overline{\text{PG}}$, STAT, $\overline{\text{INT}}$)

8.4.9.1 Power Good indicator ($\overline{\text{PG}}$ Pin and PG_STAT Bit)

The PG_STAT bit goes HIGH and $\overline{\text{PG}}$ pin goes LOW to indicate a good input source when:

- VBUS above $V_{\text{VBUS_UVLO}}$
- VBUS above battery (not in sleep)
- VBUS below $V_{\text{VAC_OV}}$ threshold
- VBUS above V_{VBUSMin} (typical 3.8 V) when I_{BADSRC} (typical 30 mA) current is applied (not a poor source)
- Completed *input Source Type Detection*

8.4.9.2 Charging Status indicator (STAT)

The device indicates charging state on the open drain STAT pin. The STAT pin can drive LED. The STAT pin function can be disabled by setting the EN_ICHG_MON bits = 11.

Table 4. STAT Pin State

CHARGING STATE	STAT INDICATOR
Charging in progress (including recharge)	LOW
Charging complete	HIGH
Sleep mode, charge disable	HIGH
Charge suspend (input overvoltage, TS fault, timer fault or system overvoltage) Boost Mode suspend (due to TS fault)	Blinking at 1 Hz

8.4.9.3 interrupt to Host ($\overline{\text{INT}}$)

in some applications, the host does not always monitor the charger operation. The INT pulse notifies the system on the device operation. The following events will generate 256- μs INT pulse.

- USB/adaptor source identified (through PSEL detection)
- Good input source detected
 - VBUS above battery (not in sleep)
 - VBUS below $V_{\text{VAC_OV}}$ threshold
 - VBUS above V_{VBUSMin} (typical 3.8 V) when I_{BADSRC} (typical 30 mA) current is applied (not a poor source)
- input removed
- Charge Complete
- Any FAULT event in REG09
- VINDPM / IINDPM event detected (maskable)

When a fault occurs, the charger device sends out INT and keeps the fault state in REG09 until the host reads the fault register. Before the host reads REG09 and all the faults are cleared, the charger device would not send any INT upon new faults. To read the current fault status, the host has to read REG09 two times consecutively. The first read reports the pre-existing fault register status and the second read reports the current fault register status.

8.4.10 Protections

8.4.10.1 Voltage and Current Monitoring in Converter Operation

The device closely monitors the input and system voltage, as well as internal FET currents for safe buck and boost mode operation.

8.4.10.1.1 Voltage and Current Monitoring in Buck Mode

8.4.10.1.1.1 input Overvoltage (ACOV)

If VBUS voltage exceeds V_{VAC_OV} (programmable via OVP[2:0] bits), the device stops switching immediately.

During input overvoltage event (ACOV), the fault register CHRG_FAULT bits are set to 01. An INT pulse is asserted to the host. The device will automatically resume normal operation once the input voltage drops back below the OVP threshold.

8.4.10.1.1.2 System Overvoltage Protection (YSOVP)

The charger device clamps the system voltage during load transient so that the components connect to system would not be damaged due to high voltage. YSOVP threshold is 350 mV above system regulation voltage. Upon YSOVP, converter stops immediately to clamp the overshoot. The charger provides 30 mA discharge current to bring down the system voltage.

8.4.10.2 Voltage and Current Monitoring in Boost Mode

The device closely monitors the VBUS voltage, as well as RBFET and LSFET current to ensure safe boost mode operation.

8.4.10.2.1 VBUS Soft Start

When the boost function is enabled, the device soft-starts boost mode to avoid inrush current.

8.4.10.2.2 VBUS Output Protection

The device monitors boost output voltage and other conditions to provide output short circuit and overvoltage protection. The Boost build in accurate constant current regulation to allow OTG to adaptive to various types of load. If short circuit is detected on VBUS, the Boost turns off and retry 7 times. If retries are not successful, OTG is disabled with OTG_CONFIG bit cleared. In addition, the BOOST_FAULT bit is set and INT pulse is generated. The BOOST_FAULT bit can be cleared by host by re-enabling boost mode

8.4.10.2.3 Boost Mode Overvoltage Protection

When the VBUS voltage rises above regulation target and exceeds VOTG_OVP, the device enters overvoltage protection which stops switching, clears OTG_CONFIG bit and exits boost mode. At Boost overvoltage duration, the fault register bit (BOOST_FAULT) is set high to indicate fault in boost operation. An INT is also asserted to the host.

8.4.10.3 Thermal Regulation and Thermal Shutdown

8.4.10.3.1 Thermal Protection in Buck Mode

The bq25601 monitors the internal junction temperature T_j to avoid overheat the chip and limits the IC surface temperature at 110°C in buck mode. When the internal junction temperature exceeds thermal regulation limit (110°C), the device lowers down the charge current. During thermal regulation, the actual charging current is usually below the programmed battery charging current. Therefore, termination is disabled, the safety timer runs at half the clock rate, and the status register THERM_STAT bit goes high.

Additionally, the device has thermal shutdown to turn off the converter and BATFET when IC surface temperature exceeds $T_{SHUT}(160^{\circ}\text{C})$. The fault register CHRG_FAULT is set to 1 and an $\overline{\text{INT}}$ is asserted to the host. The BATFET and converter is enabled to recover when IC temperature is $T_{SHUT_HYS}(30^{\circ}\text{C})$ below $T_{SHUT}(160^{\circ}\text{C})$.

8.4.10.3.2 Thermal Protection in Boost Mode

The device monitors the internal junction temperature to provide thermal shutdown during boost mode. When IC surface temperature exceeds $T_{SHUT}(160^{\circ}\text{C})$, the boost mode is disabled (converter is turned off) by setting OTG_CONFIG bit low and BATFET is turned off. When IC surface temperature is below $T_{SHUT}(160^{\circ}\text{C}) - T_{SHUT_HYS}(30^{\circ}\text{C})$, the BATFET is enabled automatically to allow system to restore and the host can re-enable OTG_CONFIG bit to recover.

8.4.10.4 Battery Protection

8.4.10.4.1 Battery overvoltage Protection (BATOVP)

The battery overvoltage limit is clamped at 4% above the battery regulation voltage. When battery over voltage occurs, the charger device immediately disables charge. The fault register BAT_FAULT bit goes high and an INT is asserted to the host.

8.4.10.4.2 Battery Over-Discharge Protection

When battery is discharged below $V_{BAT_DPL_FALL}$, the BATFET is turned off to protect battery from over discharge. To recover from over-discharge, an input source is required at VBUS. When an input source is plugged in, the BATFET turns on. The battery is charged with I_{SHORT} (typically 100 mA) current when the $\text{VBAT} < \text{VSHORT}$, or precharge current as set in IPRECHG register when the battery voltage is between V_{SHORTZ} and V_{BAT_LOWV} .

8.4.10.4.3 System Over-Current Protection

When the system is shorted or significantly overloaded ($\text{IBAT} > \text{IBATOP}$) so that its current exceeds the overcurrent limit, the device latches off BATFET. Section BATFET Enable (Exit Shipping Mode) can reset the latch-off condition and turn on BATFET.

8.4.11 Serial interface

The device uses I²C compatible interface for flexible charging parameter programming and instantaneous device status reporting. I²CTM is a bi-directional 2-wire serial interface developed by Philips Semiconductor (now NXP Semiconductors). Only two bus lines are required: a serial data line (SDA) and a serial clock line (SCL). Devices can be considered as masters or slaves when performing data transfers. A master is the device which initiates a data transfer on the bus and generates the clock signals to permit that transfer. At that time, any device addressed is considered a slave.

The device operates as a slave device with address 6BH, receiving control inputs from the master device like micro controller or a digital signal processor through REG00-REG0B. Register read beyond REG0B (0x0B) returns 0xFF. The I²C interface supports both standard mode (up to 100kbits), and fast mode (up to 400kbits). connecting to the positive supply voltage via a current source or pull-up resistor. When the bus is free, both lines are HIGH. The SDA and SCL pins are open drain.

8.4.11.1 Data Validity

The data on the SDA line must be stable during the HIGH period of the clock. The HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW. One clock pulse is generated for each data bit transferred.

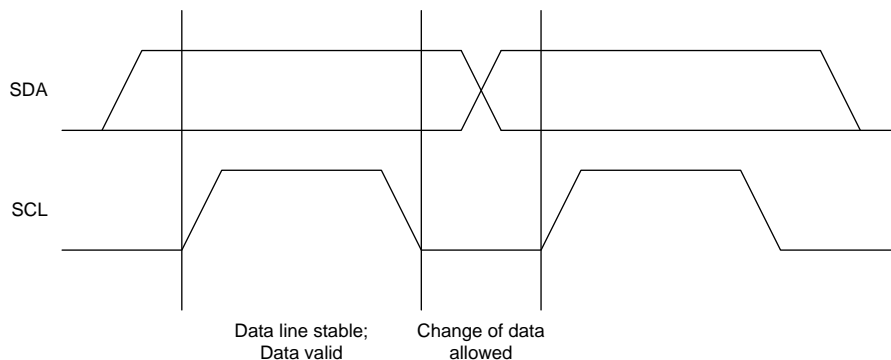


Figure 20. Bit Transfer on the I²C Bus

8.4.11.2 START and STOP Conditions

All transactions begin with a START (S) and can be terminated by a STOP (P). A HIGH to LOW transition on the SDA line while SCL is HIGH defines a START condition. A LOW to HIGH transition on the SDA line when the SCL is HIGH defines a STOP condition. START and STOP conditions are always generated by the mAsTer. The bus is considered busy after the START condition, and free after the STOP condition.

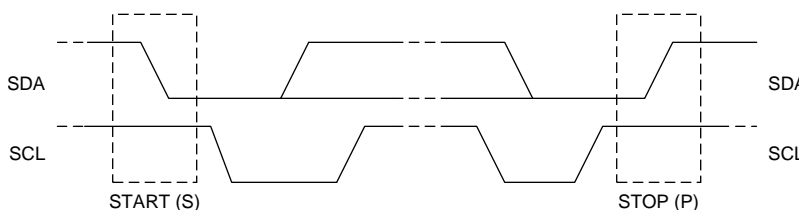


Figure 21. TS START and STOP conditions

8.4.11.3 Byte Format

Every byte on the SDA line must be 8 bits long. The number of bytes to be transmitted per transfer is unrestricted. Each byte has to be followed by an Acknowledge bit. Data is transferred with the Most Significant Bit (MSB) first. If a slave cannot receive or transmit another complete byte of data until it has performed some other function, it can hold the clock line SCL low to force the mAsTer into a wait state (clock stretching). Data transfer then continues when the slave is ready for another byte of data and release the clock line SCL.

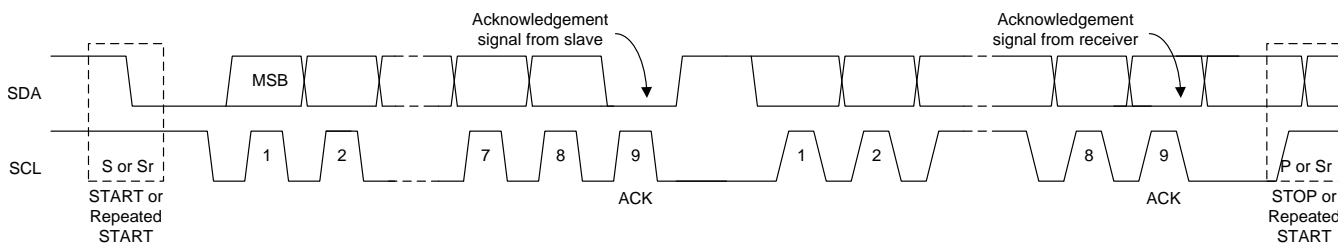


Figure 22. Data Transfer on the I²C Bus

8.4.11.4 Acknowledge (ACK) and Not Acknowledge (NACK)

The acknowledge takes place after every byte. The acknowledge bit allows the receiver to signal the transmitter that the byte was successfully received and another byte may be sent. All clock pulses, including the acknowledge ninth clock pulse, are generated by the mAsTer. The transmitter releases the SDA line during the acknowledge clock pulse so the receiver can pull the SDA line LOW and it remains stable LOW during the HIGH period of this clock pulse.

When SDA remains HIGH during the ninth clock pulse, this is the Not Acknowledge signal. The mAsTer can then generate either a STOP to abort the transfer or a repeated START to start a new transfer.

8.4.11.5 Slave Address and Data Direction Bit

After the START, a slave address is sent. This address is 7 bits long followed by the eighth bit as a data direction bit (bit R/W). A zero indicates a transmission (WRITE) and a one indicates a request for data (READ).

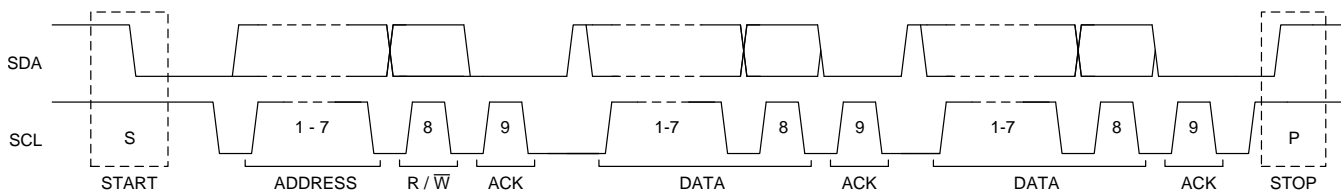


Figure 23. Complete Data Transfer

8.4.11.6 Single Read and Write

If the register address is not defined, the charger IC send back NACK and go back to the idle state.

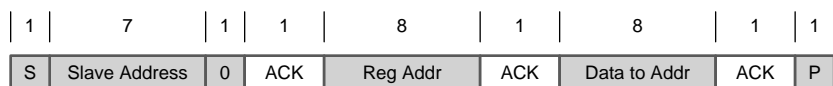


Figure 24. Single Write

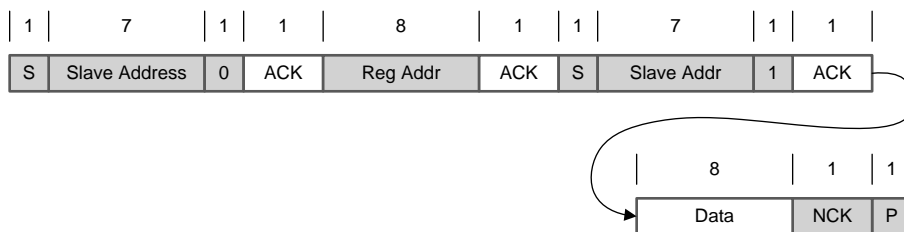


Figure 25. Single Read

8.4.11.7 Multi-Read and Multi-Write

The charger device supports multi-read and multi-write on REG00 through REG0B.

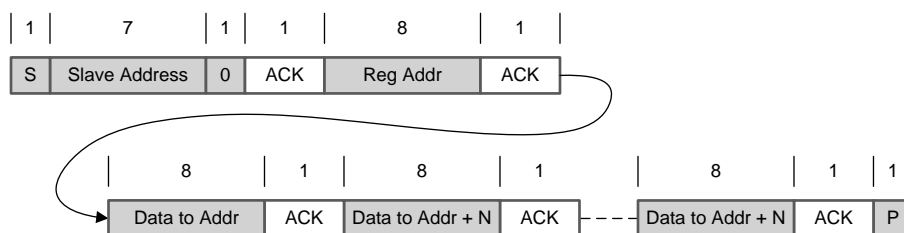


Figure 26. Multi-Write

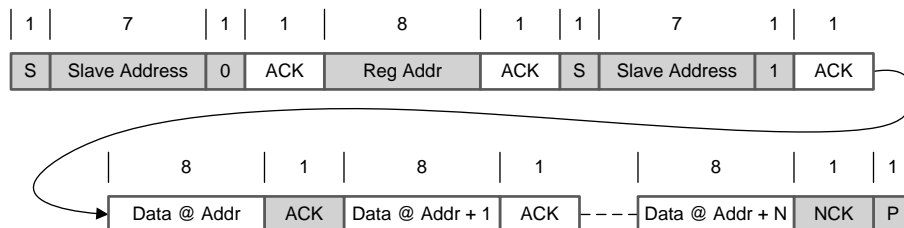


Figure 27. Multi-Read

REG09 is a fault register. It keeps all the fault information from last read until the host issues a new read. For example, if Charge Safety Timer Expiration fault occurs but recovers later, the fault register REG09 reports the fault when it is read the first time, but returns to normal when it is read the second time. In order to get the fault information at present, the host has to read REG09 for the second time. The only exception is NTC_FAULT which always reports the actual condition on the TS pin. In addition, REG09 does not support multi-read and multi-write.

8.5 Register Maps

 I²C Slave Address: 6BH

8.5.1 REG00

Table 5. REG00 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	EN_HIZ	0	R/W	by REG_RST by Watchdog	0 – Disable, 1 – Enable	Enable HIZ Mode 0 – Disable (default) 1 – Enable
6	EN_ICHG_MON[1]	0	R/W	by REG_RST	00 - Enable STAT pin function (default) 01 - Reserved 10 - Reserved 11 - Disable STAT pin function (float pin)	
5	EN_ICHG_MON[0]	0	R/W	by REG_RST		
4	IINDPM[4]	1	R/W	by REG_RST	1600 mA	Input Current Limit Offset: 100 mA Range: 100 mA (000000) – 3.2 A (11111) Default: 2400 mA (10111), maximum input current limit, not typical. IINDPM bits are changed automatically after input source detection is completed PSEL = Hi = 500 mA PSEL = Lo = 2.4 A Host can over-write IINDPM register bits after input source detection is completed.
3	IINDPM[3]	0	R/W	by REG_RST	800 mA	
2	IINDPM[2]	1	R/W	by REG_RST	400 mA	
1	IINDPM[1]	1	R/W	by REG_RST	200 mA	
0	IINDPM[0]	1	R/W	by REG_RST	100 mA	

LEGEND: R/W = Read/Write; R = Read only

8.5.2 REG01

Table 6. REG01 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	PFM_DIS	0	R/W	by REG_RST	0 – Enable PFM 1 – Disable PFM	Default: 0 - Enable
6	WD_RST	0	R/W	by REG_RST by Watchdog	I ² C Watchdog Timer Reset 0 – Normal ; 1 – Reset	Default: Normal (0) Back to 0 after watchdog timer reset
5	OTG_CONFIG	0	R/W	by REG_RST by Watchdog	0 – OTG Disable 1 – OTG Enable	Default: OTG disable (0) Note: 1. OTG_CONFIG would over-ride Charge Enable Function in CHG_CONFIG
4	CHG_CONFIG	1	R/W	by REG_RST by Watchdog	0 - Charge Disable 1- Charge Enable	Default: Charge Battery (1) Note: 1. Charge is enabled when both CE pin is pulled low AND CHG_CONFIG bit is 1.
3	SYS_Min[2]	1	R/W	by REG_RST	System Minimum Voltage	000: 2.6 V 001: 2.8 V 010: 3 V 011: 3.2 V 100: 3.4 V 101: 3.5 V 110: 3.6 V 111: 3.7 V Default: 3.5 V (101)
2	SYS_Min[1]	0	R/W	by REG_RST		
1	SYS_Min[0]	1	R/W	by REG_RST		
0	Min_VBAT_SEL	0	R/W	by REG_RST		

LEGEND: R/W = Read/Write; R = Read only

8.5.3 REG02
Table 7. REG02 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	BOOST_LIM	1	R/W	by REG_RST by Watchdog	0 = 0.5 A 1 = 1.2 A	Default: 1.2 A (1) Note: The current limit options listed are minimum current limit specs.
6	Q1_FULLON	0	R/W	by REG_RST	0 – Use higher Q1 RDSON when programmed IINDPM < 700mA (better accuracy) 1 – Use lower Q1 RDSON always (better efficiency)	In boost mode, full FET is always used and this bit has no effect
5	ICHG[5]	1	R/W	by REG_RST by Watchdog	1920 mA	Fast Charge Current Default: 2040mA (100010) Range: 0 mA (0000000) – 3000 mA (110010) Note: I _{CHG} = 0 mA disables charge. I _{CHG} > 3000 mA (110010 clamped to register value 3000 mA (110010))
4	ICHG[4]	0	R/W	by REG_RST by Watchdog	960 mA	
3	ICHG[3]	0	R/W	by REG_RST by Watchdog	480 mA	
2	ICHG[2]	0	R/W	by REG_RST by Watchdog	240 mA	
1	ICHG[1]	1	R/W	by REG_RST by Watchdog	120 mA	
0	ICHG[0]	0	R/W	by REG_RST by Watchdog	60 mA	

LEGEND: R/W = Read/Write; R = Read only

8.5.4 REG03

Table 8. REG03 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	IPRECHG[3]	0	R/W	by REG_RST by Watchdog	480 mA	Precharge Current Default: 180 mA (0010) Offset: 60 mA Note: IPRECHG > 780 mA clamped to 780 mA (1100)
6	IPRECHG[2]	0	R/W	by REG_RST by Watchdog	240 mA	
5	IPRECHG[1]	1	R/W	by REG_RST by Watchdog	120 mA	
4	IPRECHG[0]	0	R/W	by REG_RST by Watchdog	60 mA	
3	ITERM[3]	0	R/W	by REG_RST by Watchdog	480 mA	Termination Current Default: 180 mA (0010) Offset: 60 mA
2	ITERM[2]	0	R/W	by REG_RST by Watchdog	240 mA	
1	ITERM[1]	1	R/W	by REG_RST by Watchdog	120 mA	
0	ITERM[0]	0	R/W	by REG_RST by Watchdog	60 mA	

LEGEND: R/W = Read/Write; R = Read only

8.5.5 REG04
Table 9. REG04 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	VREG[4]	0	R/W	by REG_RST by Watchdog	512 mV	Charge Voltage Offset: 3.856 V Range: 3.856 V to 4.624 V (11000) Default: 4.208 V (01011) Special Value: (01111): 4.352 V Note: Value above 11000 (4.624 V) is clamped to register value 11000 (4.624 V)
6	VREG[3]	1	R/W	by REG_RST by Watchdog	256 mV	
5	VREG[2]	0	R/W	by REG_RST by Watchdog	128 mV	
4	VREG[1]	1	R/W	by REG_RST by Watchdog	64 mV	
3	VREG[0]	1	R/W	by REG_RST by Watchdog	32 mV	
2	TOPOFF_TIMER[1]	0	R/W	by REG_RST by Watchdog	00 – Disabled (Default) 01 – 15 minutes	The extended time following the termination condition is met. When disabled, charge terminated when termination conditions are met
1	TOPOFF_TIMER[0]	0	R/W	by REG_RST by Watchdog	10 – 30 minutes 11 – 45 minutes	
0	VRECHG	0	R/W	by REG_RST by Watchdog	0 – 100 mV 1 – 200 mV	Recharge threshold Default: 100mV (0)

LEGEND: R/W = Read/Write; R = Read only

8.5.6 REG05

Table 10. REG05 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	EN_TERM	1	R/W	by REG_RST by Watchdog	0 – Disable 1 – Enable	Default: Enable termination (1)
6	Reserved	0	R/W	by REG_RST by Watchdog	Reserved	Reserved
5	WATCHDOG[1]	0	R/W	by REG_RST by Watchdog	00 – Disable timer, 01 – 40 s, 10 – 80 s, 11 – 160 s	Default: 40 s (01)
4	WATCHDOG[0]	1	R/W	by REG_RST by Watchdog		
3	EN_TIMER	1	R/W	by REG_RST by Watchdog	0 – Disable 1 – Enable both fast charge and precharge timer	Default: Enable (1)
2	CHG_TIMER	1	R/W	by REG_RST by Watchdog	0 – 5 hrs 1 – 10 hrs	Default: 10 hours (1)
1	TREG	1	R/W	by REG_RST by Watchdog	Thermal Regulation Threshold: 0 - 90°C 1 - 110°C	Default: 110°C (1)
0	JEITA_ISET (0C-10C)	1	R/W	by REG_RST by Watchdog	0 – 50% of ICHG 1 – 20% of ICHG	Default: 20% (1)

LEGEND: R/W = Read/Write; R = Read only

8.5.7 REG06
Table 11. REG06 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	OVP[1]	0	R/W	by REG_RST	Default: 6.5V (01)	VAC OVP threshold: 00 - 5.5 V 01 – 6.5 V (5-V input) 10 – 10.5 V (9-V input) 11 – 14 V (12-V input)
6	OVP[0]	1	R/W	by REG_RST		
5	BOOSTV[1]	1	R/W	by REG_RST		Boost Regulation Voltage: 00 - 4.85V 01 - 5.00V 10 - 5.15V 11 - 5.30V
4	BOOSTV[0]	0	R/W	by REG_RST		
3	VINDPM[3]	0	R/W	by REG_RST	800 mV	Absolute VINDPM Threshold Offset: 3.9 V Range: 3.9 V (0000) – 5.4 V (1111) Default: 4.5V (0110)
2	VINDPM[2]	1	R/W	by REG_RST	400 mV	
1	VINDPM[1]	1	R/W	by REG_RST	200 mV	
0	VINDPM[0]	0	R/W	by REG_RST	100 mV	

LEGEND: R/W = Read/Write; R = Read only

8.5.8 REG07

Table 12. REG07 Field Descriptions

Bit	Field	POR	Type	Reset	Description	Comment
7	IINDET_EN	0	R/W	by REG_RST by Watchdog	0 - Not in input current limit detection 1 - Force input current limit detection when VBUS is present	Returns to 0 after input detection is complete
6	TMR2X_EN	1	R/W	by REG_RST by Watchdog	0 – Disable 1 – Safety timer slowed by 2X during input DPM (both V and I) or JEITA cool, or thermal regulation	
5	BATFET_DIS	0	R/W	by REG_RST	0 – Allow Q4 turn on, 1 – Turn off Q4 with $t_{\text{BATFET_DLY}}$ delay time (REG07[3])	Default: Allow Q4 turn on(0)
4	JEITA_VSET (45C-60C)	0	R/W	by REG_RST by Watchdog	0 – Set Charge Voltage to 4.1V (max), 1 – Set Charge Voltage to VREG	
3	BATFET_DLY	1	R/W	by REG_RST	0 – Turn off BATFET immediately when BATFET_DIS bit is set 1 – Turn off BATFET after $t_{\text{BATFET_DLY}}$ (typ. 10 s) when BATFET_DIS bit is set	Default: 1 Turn off BATFET after $t_{\text{BATFET_DLY}}$ (typ. 10 s) when BATFET_DIS bit is set
2	BATFET_RST_EN	1	R/W	by REG_RST by Watchdog	0 – Disable BATFET reset function 1 – Enable BATFET reset function	Default: 1 Enable BATFET reset function
1	VDPM_BAT_TRACK[1]	0	R/W	by REG_RST	00 - Disable function (VINDPM set by register) 01 - VBAT + 200mV 10 - VBAT + 250mV 11 - VBAT + 300mV	Sets VINDPM to track BAT voltage. Actual VINDPM is higher of register value and VBAT + VDPM_BAT_TRACK
0	VDPM_BAT_TRACK[0]	0	R/W	by REG_RST		

LEGEND: R/W = Read/Write; R = Read only

8.5.9 REG08
Table 13. REG08 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	VBUS_STAT[2]	x	R	NA	VBUS Status register 000: No input 001: USB Host SDP (500 mA) → PSEL HIGH 010: Adapter 2.4A → PSEL LOW 111: OTG Software current limit is reported in IINDPM register
6	VBUS_STAT[1]	x	R	NA	
5	VBUS_STAT[0]	x	R	NA	
4	CHRG_STAT[1]	x	R	NA	Charging status: 00 – Not Charging 01 – Pre-charge ($< V_{BATLOWV}$) 10 – Fast Charging 11 – Charge Termination
3	CHRG_STAT[0]	x	R	NA	
2	PG_STAT	x	R	NA	Power Good status: 0 – Power Not Good 1 – Power Good
1	THERM_STAT	x	R	NA	0 – Not in ther mA1 regulation 1 – in ther mA1 regulation
0	VSYS_STAT	x	R	NA	0 – Not in VSYSMin regulation (BAT > VSYSMin) 1 – in VSYSMin regulation (BAT < VSYSMin)

LEGEND: R/W = Read/Write

8.5.10 REG09**Table 14. REG09 Field Descriptions**

Bit	Field	POR	Type	Reset	Description
7	WATCHDOG_FAULT	x	R	NA	0 – Normal, 1- Watchdog timer expiration
6	BOOST_FAULT	x	R	NA	0 – Normal, 1 – VBUS overloaded in OTG, or VBUS OVP, or battery is too low (any conditions that we cannot start boost function)
5	CHRG_FAULT[1]	x	R	NA	00 – Normal, 01 – input fault (VAC OVP or VBAT < VBUS < 3.8 V), 10 - Thermal shutdown, 11 – Charge Safety Timer Expiration
4	CHRG_FAULT[0]	x	R	NA	
3	BAT_FAULT	x	R	NA	0 – Normal, 1 – BATOVP
2	NTC_FAULT[2]	x	R	NA	JEITA 000 – Normal, 010 – Warm, 011 – Cool, 101 – Cold, 110 – Hot (Buck mode)
1	NTC_FAULT[1]	x	R	NA	
0	NTC_FAULT[0]	x	R	NA	

LEGEND: R/W = Read/Write; R = Read only

8.5.11 REG0A
Table 15. REG0A Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	VBUS_GD	x	R	NA	0 – Not VBUS attached, 1 – VBUS Attached
6	VINDPM_STAT	x	R	NA	0 – Not in VINDPM, 1 – in VINDPM
5	IINDPM_STAT	x	R	NA	0 – Not in IINDPM, 1 – in IINDPM
4	Reserved	x	R	NA	
3	TOPOFF_ACTIVE	x	R	NA	0 – Top off timer not counting. 1 – Top off timer counting
2	ACOV_STAT	x	R	NA	0 – Device is NOT in ACOV 1 – Device is in ACOV
1	VINDPM_INT_MASK	0	R/W	by REG_RST	0 - Allow VINDPM INT pulse 1 - Mask VINDPM INT pulse
0	IINDPM_INT_MASK	0	R/W	by REG_RST	0 - Allow IINDPM INT pulse 1 - Mask IINDPM INT pulse

LEGEND: R/W = Read/Write; R = Read only

8.5.12 REG0B

Table 16. REG0B Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	REG_RST	0	R/W	NA	Register reset 0 – Keep current register setting 1 – Reset to default register value and reset safety timer Note: Bit resets to 0 after register reset is completed
6	PN[3]	0	R	NA	bq25601 : 0010
5	PN[2]	0	R	NA	
4	PN[1]	1	R	NA	
3	PN[0]	0	R	NA	
2	Reserved	x	R	NA	
1	DEV_REV[1]	x	R	NA	
0	DEV_REV[0]	x	R	NA	

LEGEND: R/W = Read/Write; R = Read only

9 Application and Implementation

NOTE

information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application information

A typical application consists of the device configured as an I²C controlled power path management device and a single cell battery charger for Li-Ion and Li-polymer batteries used in a wide range of smart phones and other portable devices. It integrates an input reverse-block FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET Q4) between the system and battery. The device also integrates a bootstrap diode for the high-side gate drive.

9.2 Typical Application Diagram

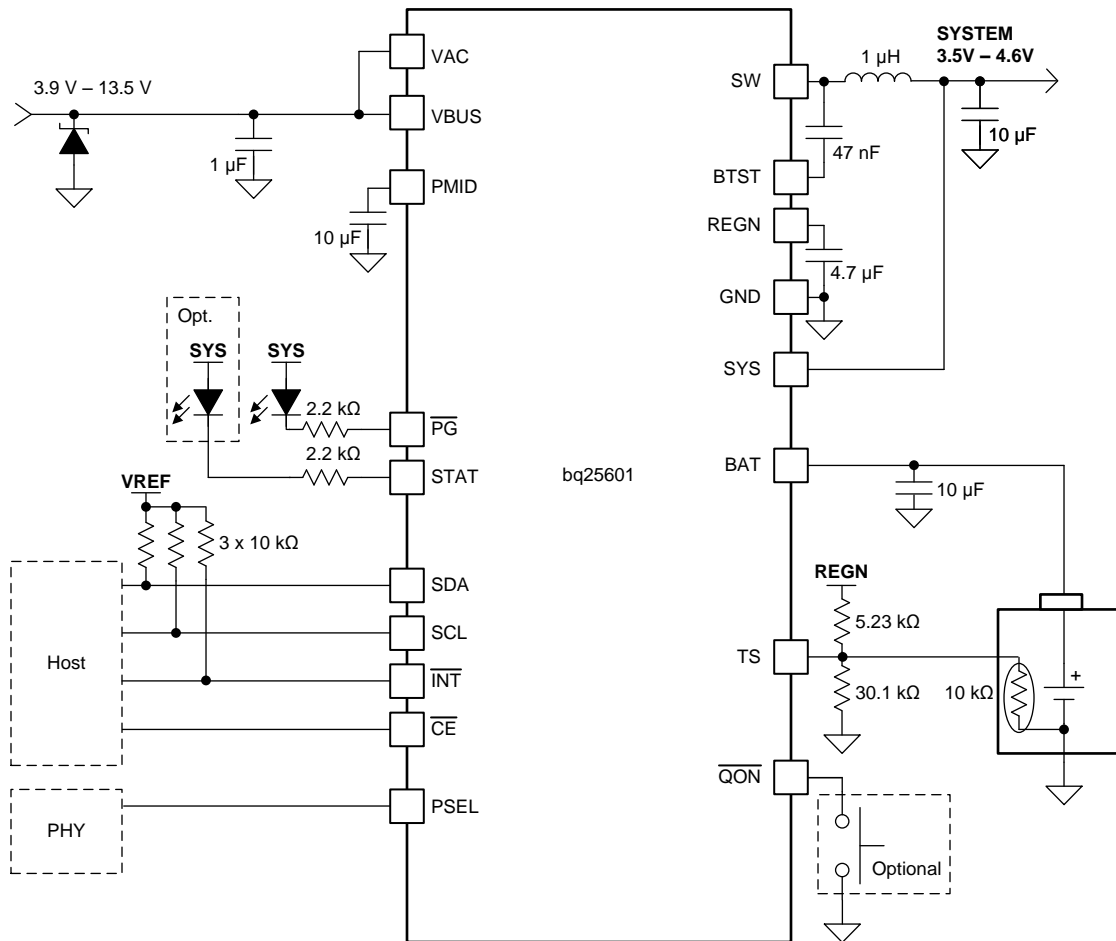


Figure 28. Typical Application

9.2.1 Design Requirements

9.2.2 Detailed Design Procedure

9.2.2.1 inductor Selection

The 1.5-MHz switching frequency allows the use of small inductor and capacitor values. maintain an inductor saturation current higher than the charging current (I_{CHG}) plus half the ripple current (I_{RIPPLE}):

$$I_{SAT} \geq I_{CHG} + (1/2) I_{RIPPLE} \quad (3)$$

The inductor ripple current depends on the input voltage (V_{VBUS}), the duty cycle ($D = V_{BAT}/V_{VBUS}$), the switching frequency (f_s) and the inductance (L).

$$I_{RIPPLE} = \frac{V_{IN} \times D \times (1 - D)}{f_s \times L} \quad (4)$$

The maximum inductor ripple current occurs when the duty cycle (D) is 0.5 or approximately 0.5. Usually inductor ripple is designed in the range between 20% and 40% maximum charging current as a trade-off between inductor size and efficiency for a practical design.

Typical Application Diagram (continued)

9.2.2.2 Input Capacitor

Design input capacitance to provide enough ripple current rating to absorb input switching ripple current. The worst case RMS ripple current is half of the charging current when duty cycle is 0.5. If the converter does not operate at 50% duty cycle, then the worst case capacitor RMS current I_{CIN} occurs where the duty cycle is closest to 50% and can be estimated using [Equation 5](#).

$$I_{CIN} = I_{CHG} \times \sqrt{D \times (1 - D)} \quad (5)$$

Low ESR ceramic capacitor such as X7R or X5R is preferred for input decoupling capacitor and should be placed to the drain of the high-side MOSFET and source of the low-side MOSFET as close as possible. Voltage rating of the capacitor must be higher than normal input voltage level. A rating of 25-V or higher capacitor is preferred for 15 V input voltage. Capacitance of 22- μ F is suggested for typical of 3A to 4 A charging current.

9.2.2.3 Output Capacitor

Ensure that the output capacitance has enough ripple current rating to absorb the output switching ripple current. [Equation 6](#) shows the output capacitor RMS current I_{COUT} calculation.

$$I_{COUT} = \frac{I_{RIPPLE}}{2 \times \sqrt{3}} \approx 0.29 \times I_{RIPPLE} \quad (6)$$

The output capacitor voltage ripple can be calculated as follows:

$$\Delta V_O = \frac{V_{OUT}}{8LCfs^2} \left(1 - \frac{V_{OUT}}{V_{IN}} \right) \quad (7)$$

At certain input and output voltage and switching frequency, the voltage ripple can be reduced by increasing the output filter LC.

The charger device has internal loop compensation optimized for >20 μ F ceramic output capacitance. The preferred ceramic capacitor is 10V rating, X7R or X5R.

9.3 Application Curves

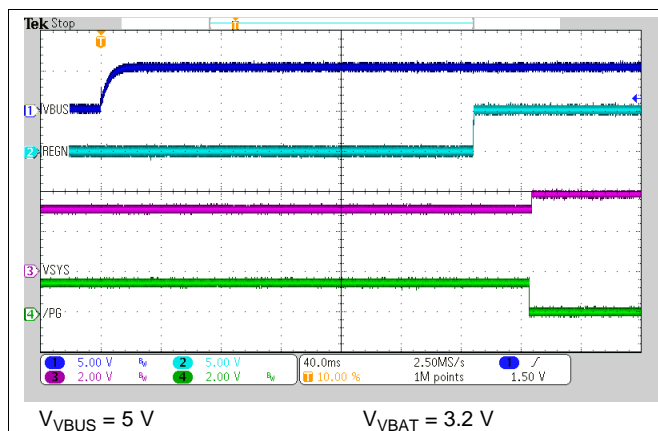


Figure 29. Power-Up with Charge Disabled

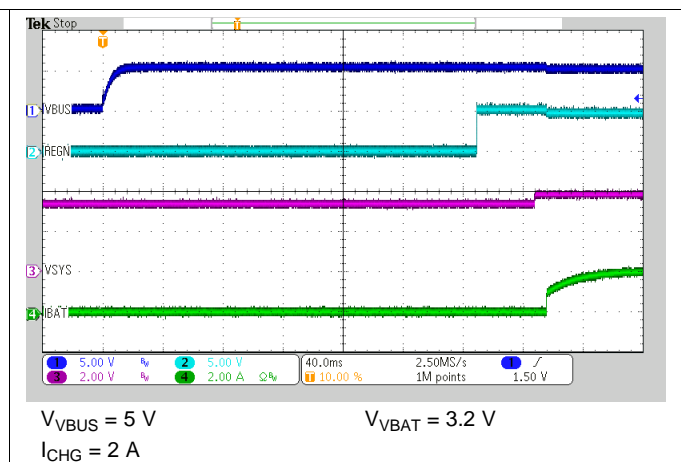


Figure 30. Power-Up with Charge Enabled

Application Curves (continued)

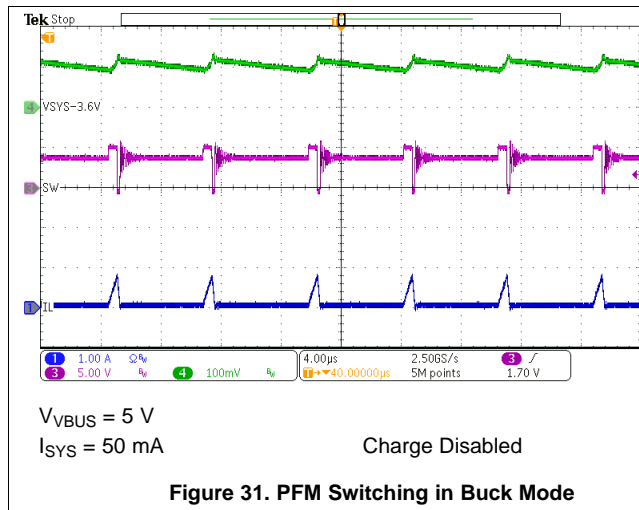


Figure 31. PFM Switching in Buck Mode

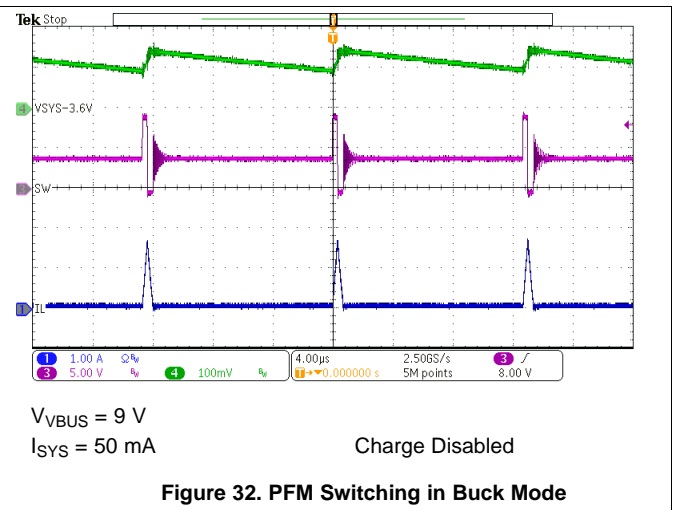


Figure 32. PFM Switching in Buck Mode

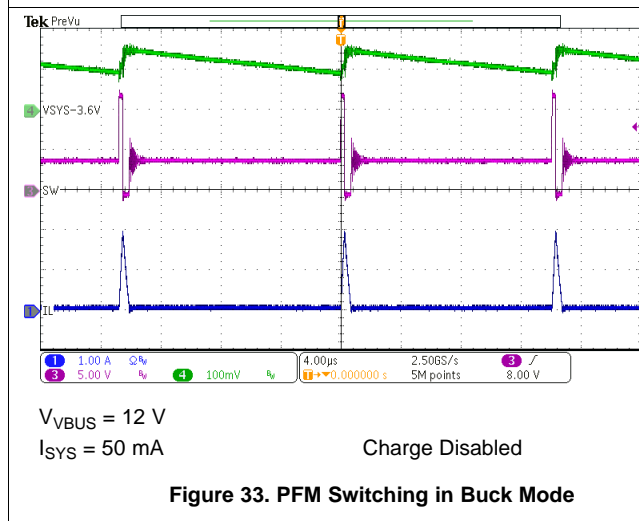


Figure 33. PFM Switching in Buck Mode

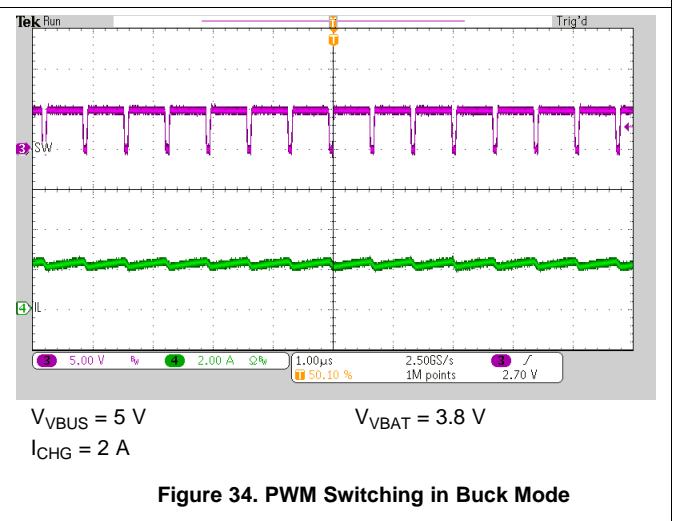


Figure 34. PWM Switching in Buck Mode

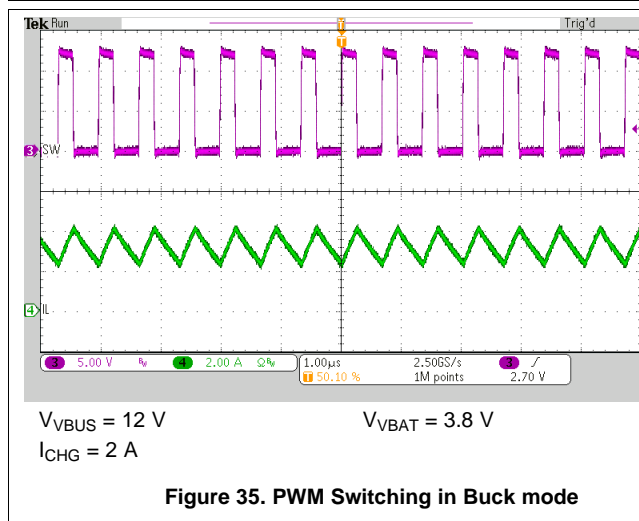


Figure 35. PWM Switching in Buck mode

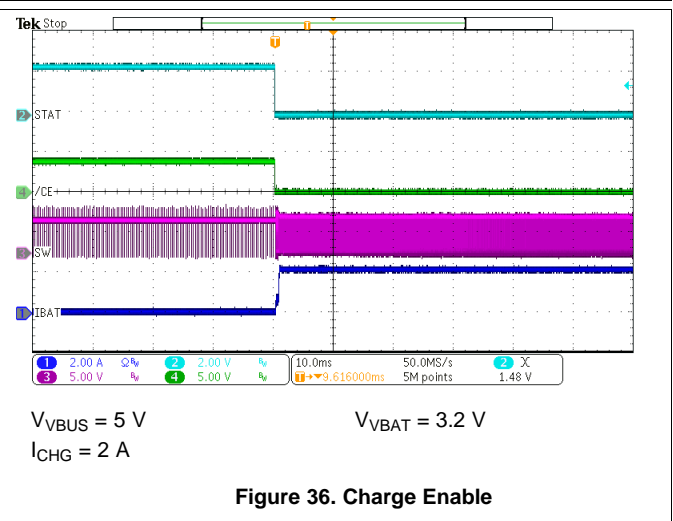


Figure 36. Charge Enable

Application Curves (continued)

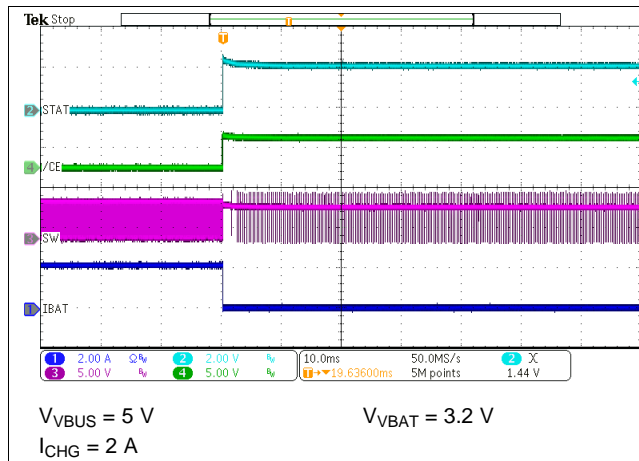


Figure 37. Charge Disable

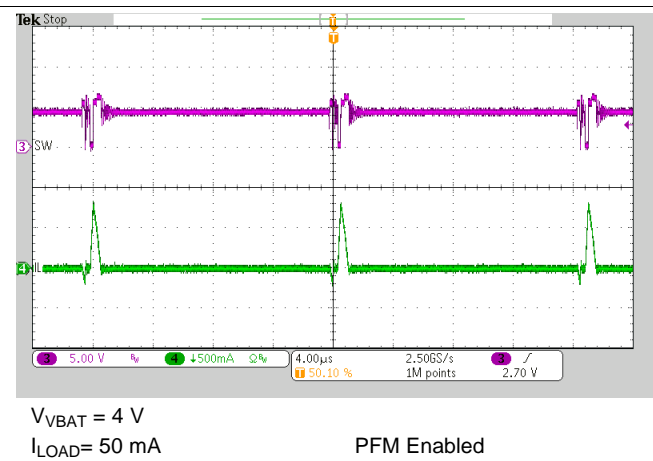


Figure 38. OTG Switching

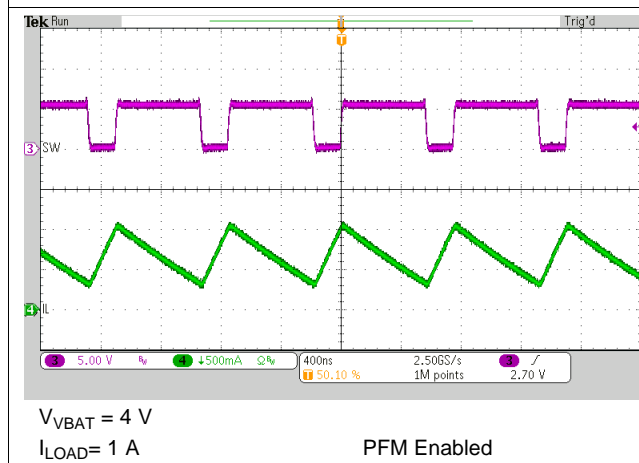


Figure 39. OTG Switching

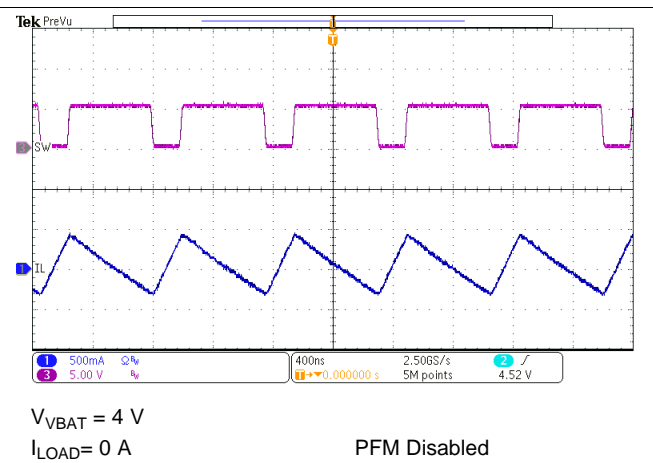


Figure 40. OTG Switching

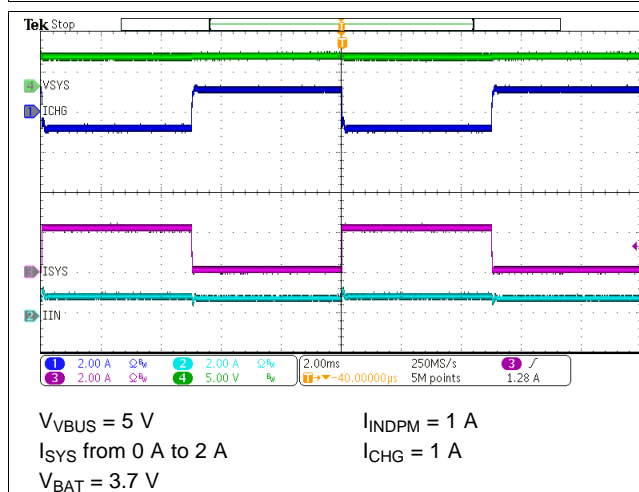


Figure 41. System Load Transient

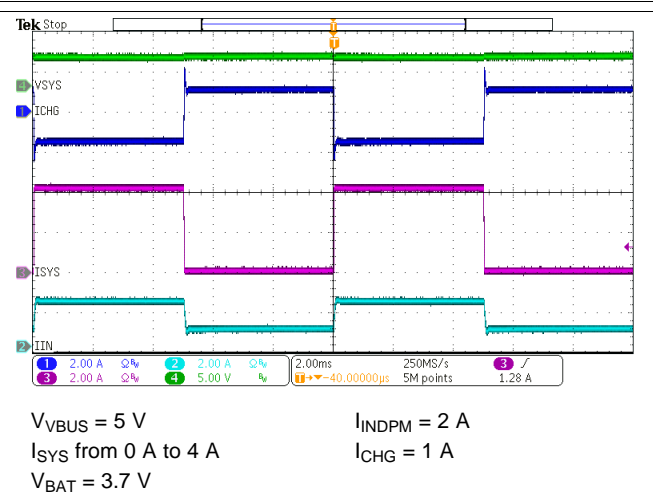


Figure 42. System Load Transient

Application Curves (continued)

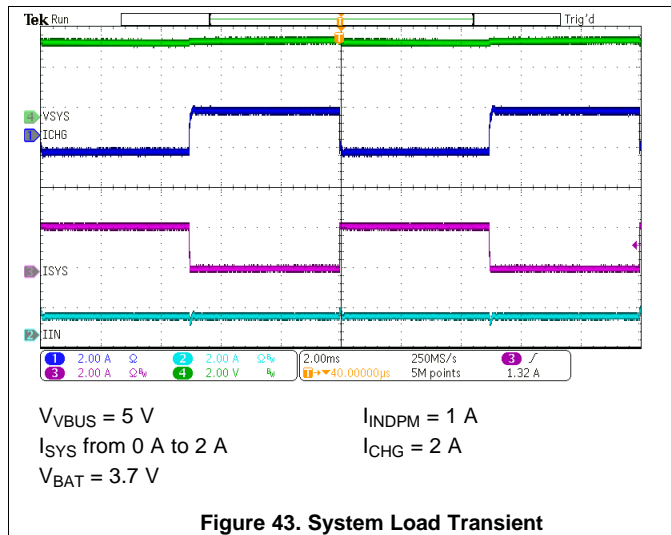


Figure 43. System Load Transient

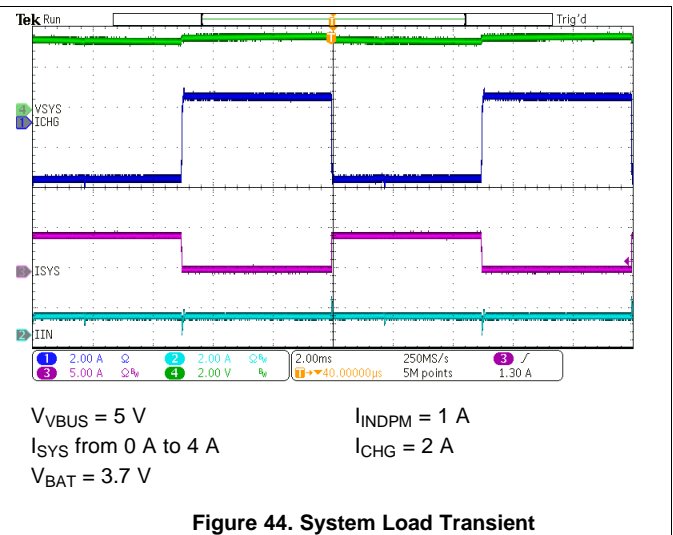


Figure 44. System Load Transient

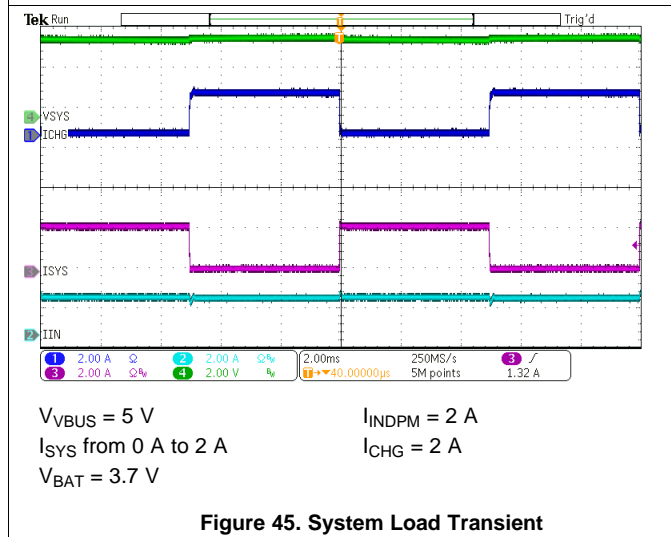


Figure 45. System Load Transient

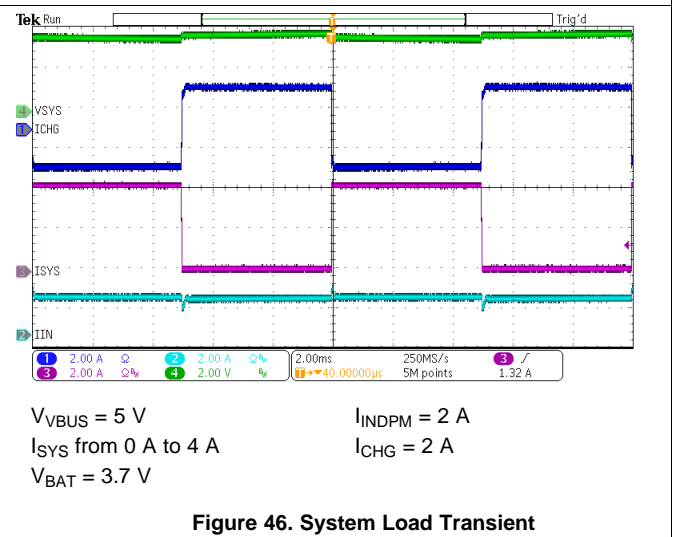


Figure 46. System Load Transient

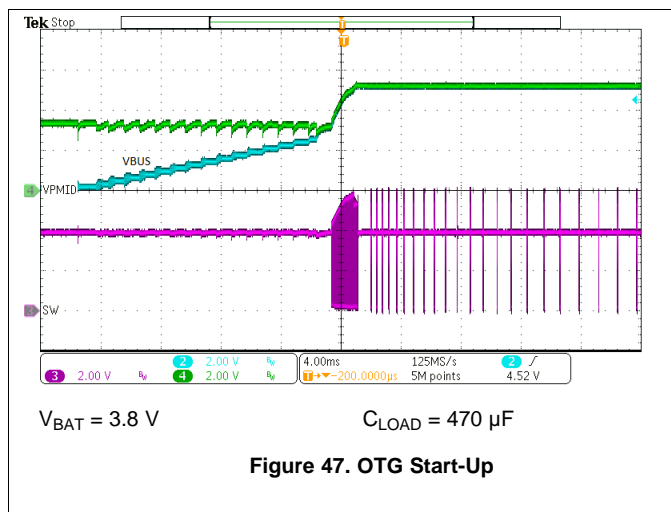


Figure 47. OTG Start-Up

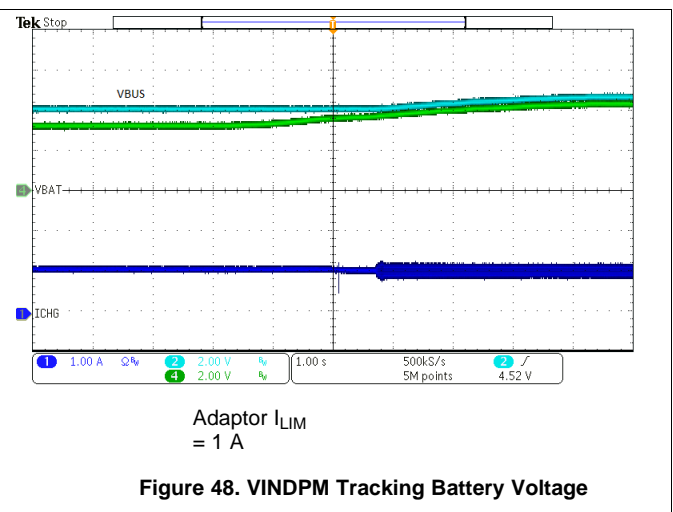


Figure 48. VINDPM Tracking Battery Voltage

10 Power Supply Recommendations

in order to provide an output voltage on SYS, the bq25601 device requires a power supply between 3.9 V and 14.2 V input with at least 100-mA current rating connected to VBUS; or, a single-cell Li-Ion battery with voltage > $V_{BATUVLO}$ connected to BAT. The source current rating needs to be at least 3 A in order for the buck converter of the charger to provide maximum output power to SYS.

11 Layout

11.1 Layout Guidelines

The switching node rise and fall times should be minimized for minimum switching loss. Proper layout of the components to minimize high frequency current path loop (see [Figure 49](#)) is important to prevent electrical and magnetic field radiation and high frequency resonant problems. Follow this specific order carefully to achieve the proper layout.

1. Place input capacitor as close as possible to PMID pin and GND pin connections and use shortest copper trace connection or GND plane.
2. Place inductor input pin to SW pin as close as possible. Minimize the copper area of this trace to lower electrical and magnetic field radiation but make the trace wide enough to carry the charging current. Do not use multiple layers in parallel for this connection. Minimize parasitic capacitance from this area to any other trace or plane.
3. Put output capacitor near to the inductor and the device. Ground connections need to be tied to the IC ground with a short copper trace connection or GND plane.
4. Route analog ground separately from power ground. Connect analog ground and connect power ground separately. Connect analog ground and power ground together using thermal pad as the single ground connection point. Or using a 0-Ω resistor to tie analog ground to power ground.
5. Use single ground connection to tie charger power ground to charger analog ground. Just beneath the device. Use ground copper pour but avoid power pins to reduce inductive and capacitive noise coupling.
6. Place decoupling capacitors next to the IC pins and make trace connection as short as possible.
7. It is critical that the exposed thermal pad on the backside of the device package be soldered to the PCB ground. Ensure that there are sufficient thermal vias directly under the IC, connecting to the ground plane on the other layers.
8. Ensure that the number and sizes of vias allow enough copper for a given current path.

See the EVM user's guide [SLUUBL4](#) for the recommended component placement with trace and via locations. For the VQFN information, refer to [SCBA017](#) and [SLUA271](#).

11.2 Layout Example

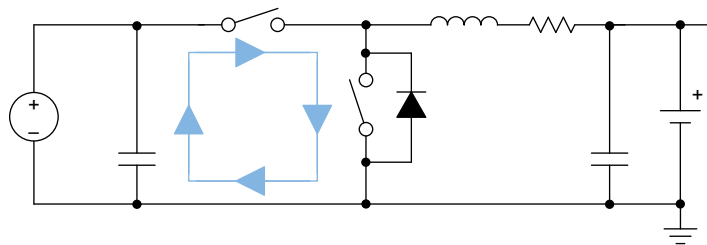


Figure 49. High Frequency Current Path

Layout Example (continued)

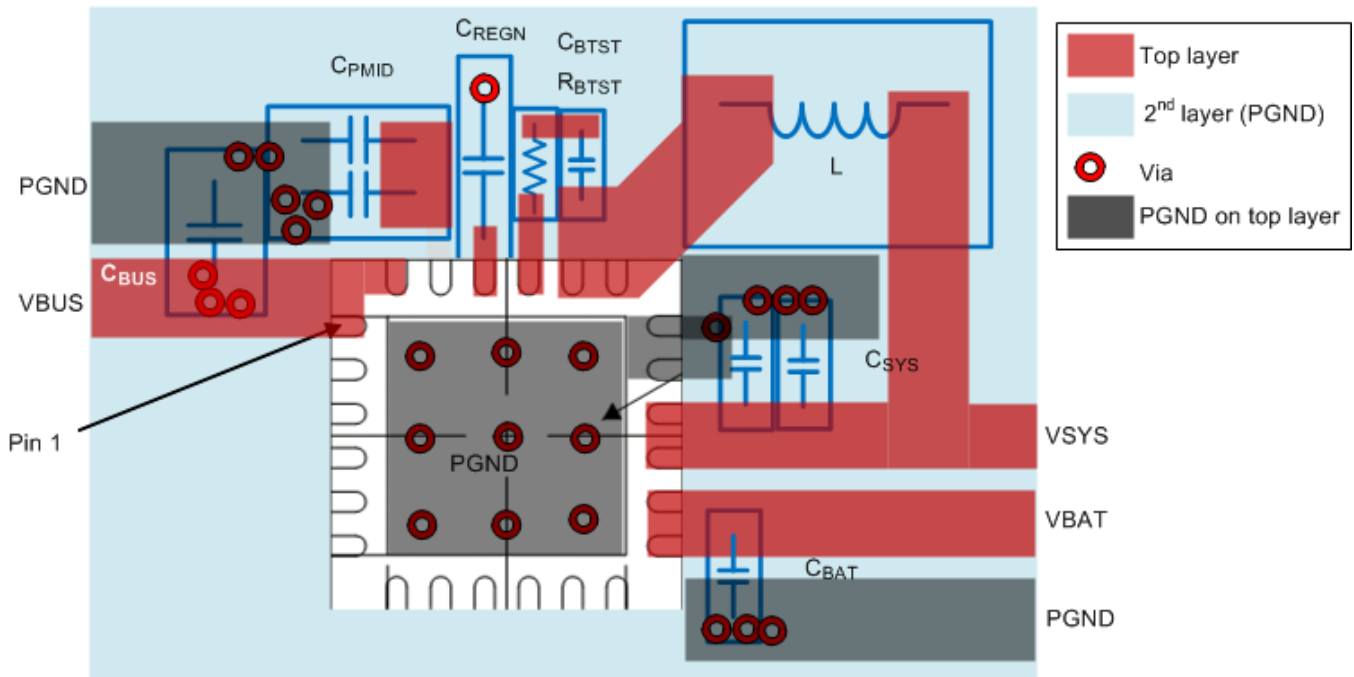


Figure 50. Layout Example

12 Device and Documentation Support

12.1 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

TI E2E™ Online Community *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.2 Trademarks

E2E is a trademark of Texas Instruments.

12.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.4 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
BQ25601RTWR	ACTIVE	WQFN	RTW	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BQ25601	Samples
BQ25601RTWT	ACTIVE	WQFN	RTW	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	BQ25601	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

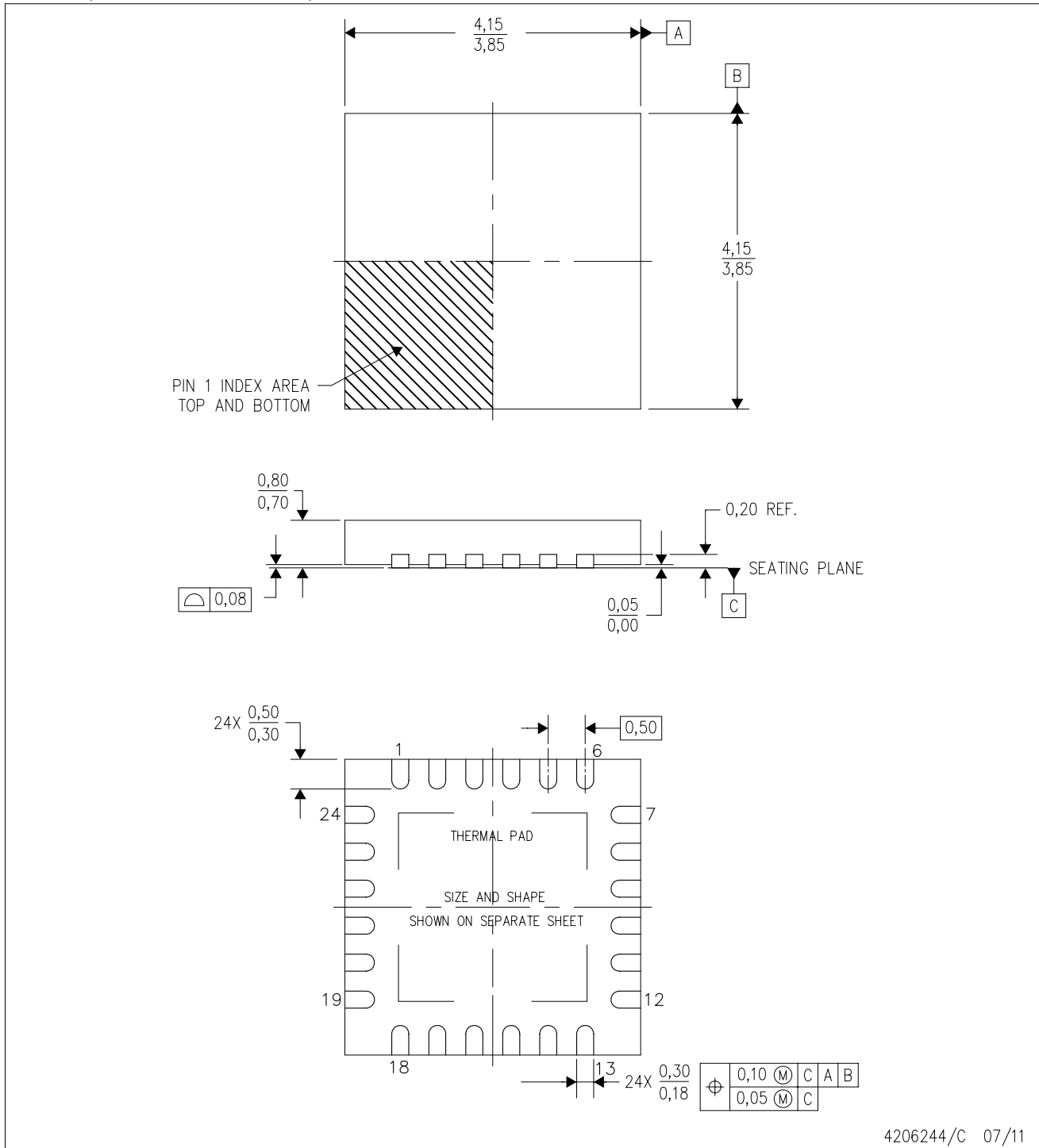
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



4206244/C 07/11

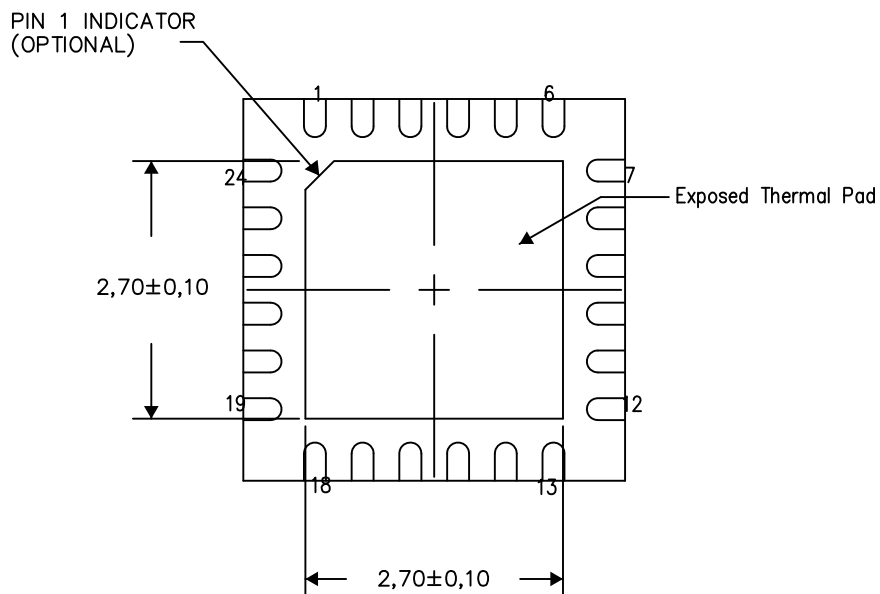
- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Quad Flatpack, No-Leads (QFN) package configuration.
 - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
 - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - F. Falls within JEDEC MO-220.

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

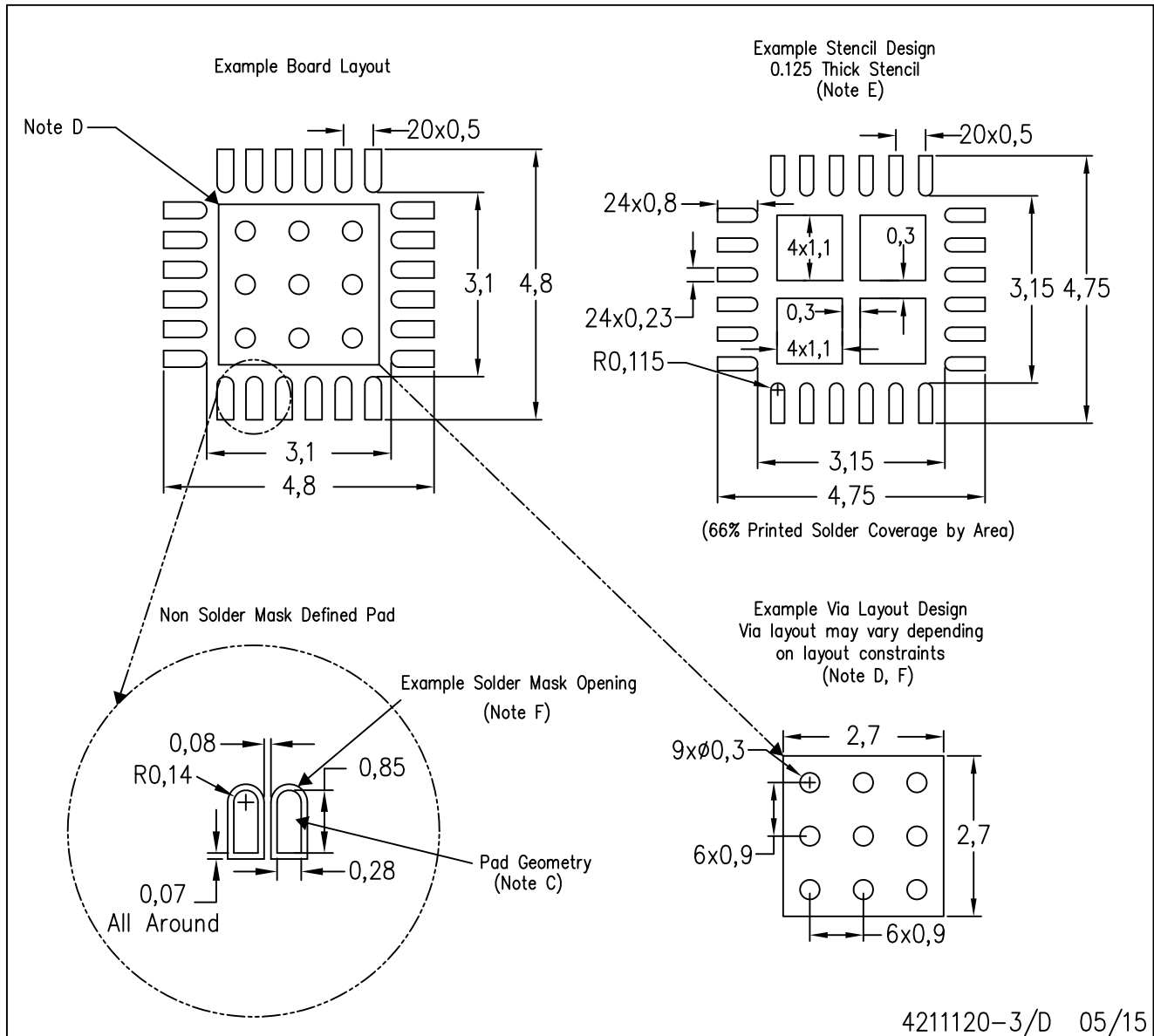
Exposed Thermal Pad Dimensions

4206249-5/P 05/15

NOTES: A. All linear dimensions are in millimeters

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.

IMPORTANT NOTICE

Texas Instruments Incorporated (TI) reserves the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete.

TI's published terms of sale for semiconductor products (<http://www.ti.com/sc/docs/stdterms.htm>) apply to the sale of packaged integrated circuit products that TI has qualified and released to market. Additional terms may apply to the use or sale of other types of TI products and services.

Reproduction of significant portions of TI information in TI data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such reproduced documentation. Information of third parties may be subject to additional restrictions. Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyers and others who are developing systems that incorporate TI products (collectively, "Designers") understand and agree that Designers remain responsible for using their independent analysis, evaluation and judgment in designing their applications and that Designers have full and exclusive responsibility to assure the safety of Designers' applications and compliance of their applications (and of all TI products used in or for Designers' applications) with all applicable regulations, laws and other applicable requirements. Designer represents that, with respect to their applications, Designer has all the necessary expertise to create and implement safeguards that (1) anticipate dangerous consequences of failures, (2) monitor failures and their consequences, and (3) lessen the likelihood of failures that might cause harm and take appropriate actions. Designer agrees that prior to using or distributing any applications that include TI products, Designer will thoroughly test such applications and the functionality of such TI products as used in such applications.

TI's provision of technical, application or other design advice, quality characterization, reliability data or other services or information, including, but not limited to, reference designs and materials relating to evaluation modules, (collectively, "TI Resources") are intended to assist designers who are developing applications that incorporate TI products; by downloading, accessing or using TI Resources in any way, Designer (individually or, if Designer is acting on behalf of a company, Designer's company) agrees to use any particular TI Resource solely for this purpose and subject to the terms of this Notice.

TI's provision of TI Resources does not expand or otherwise alter TI's applicable published warranties or warranty disclaimers for TI products, and no additional obligations or liabilities arise from TI providing such TI Resources. TI reserves the right to make corrections, enhancements, improvements and other changes to its TI Resources. TI has not conducted any testing other than that specifically described in the published documentation for a particular TI Resource.

Designer is authorized to use, copy and modify any individual TI Resource only in connection with the development of applications that include the TI product(s) identified in such TI Resource. NO OTHER LICENSE, EXPRESS OR IMPLIED, BY ESTOPPEL OR OTHERWISE TO ANY OTHER TI INTELLECTUAL PROPERTY RIGHT, AND NO LICENSE TO ANY TECHNOLOGY OR INTELLECTUAL PROPERTY RIGHT OF TI OR ANY THIRD PARTY IS GRANTED HEREIN, including but not limited to any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information regarding or referencing third-party products or services does not constitute a license to use such products or services, or a warranty or endorsement thereof. Use of TI Resources may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

TI RESOURCES ARE PROVIDED "AS IS" AND WITH ALL FAULTS. TI DISCLAIMS ALL OTHER WARRANTIES OR REPRESENTATIONS, EXPRESS OR IMPLIED, REGARDING RESOURCES OR USE THEREOF, INCLUDING BUT NOT LIMITED TO ACCURACY OR COMPLETENESS, TITLE, ANY EPIDEMIC FAILURE WARRANTY AND ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, AND NON-INFRINGEMENT OF ANY THIRD PARTY INTELLECTUAL PROPERTY RIGHTS. TI SHALL NOT BE LIABLE FOR AND SHALL NOT DEFEND OR INDEMNIFY DESIGNER AGAINST ANY CLAIM, INCLUDING BUT NOT LIMITED TO ANY INFRINGEMENT CLAIM THAT RELATES TO OR IS BASED ON ANY COMBINATION OF PRODUCTS EVEN IF DESCRIBED IN TI RESOURCES OR OTHERWISE. IN NO EVENT SHALL TI BE LIABLE FOR ANY ACTUAL, DIRECT, SPECIAL, COLLATERAL, INDIRECT, PUNITIVE, INCIDENTAL, CONSEQUENTIAL OR EXEMPLARY DAMAGES IN CONNECTION WITH OR ARISING OUT OF TI RESOURCES OR USE THEREOF, AND REGARDLESS OF WHETHER TI HAS BEEN ADVISED OF THE POSSIBILITY OF SUCH DAMAGES.

Unless TI has explicitly designated an individual product as meeting the requirements of a particular industry standard (e.g., ISO/TS 16949 and ISO 26262), TI is not responsible for any failure to meet such industry standard requirements.

Where TI specifically promotes products as facilitating functional safety or as compliant with industry functional safety standards, such products are intended to help enable customers to design and create their own applications that meet applicable functional safety standards and requirements. Using products in an application does not by itself establish any safety features in the application. Designers must ensure compliance with safety-related requirements and standards applicable to their applications. Designer may not use any TI products in life-critical medical equipment unless authorized officers of the parties have executed a special contract specifically governing such use. Life-critical medical equipment is medical equipment where failure of such equipment would cause serious bodily injury or death (e.g., life support, pacemakers, defibrillators, heart pumps, neurostimulators, and implantables). Such equipment includes, without limitation, all medical devices identified by the U.S. Food and Drug Administration as Class III devices and equivalent classifications outside the U.S.

TI may expressly designate certain products as completing a particular qualification (e.g., Q100, Military Grade, or Enhanced Product). Designers agree that it has the necessary expertise to select the product with the appropriate qualification designation for their applications and that proper product selection is at Designers' own risk. Designers are solely responsible for compliance with all legal and regulatory requirements in connection with such selection.

Designer will fully indemnify TI and its representatives against any damages, costs, losses, and/or liabilities arising out of Designer's non-compliance with the terms and provisions of this Notice.